

EE331: Day 1: Syllabus, Overview, Modeling Time

Syllabus notes:

- Labs: 25%
 - HW: 10%
 - Quizzes: 5%
 - Exams: 60%
- MT (Best) 20%
MT (Worst) 10%
Final: 30%

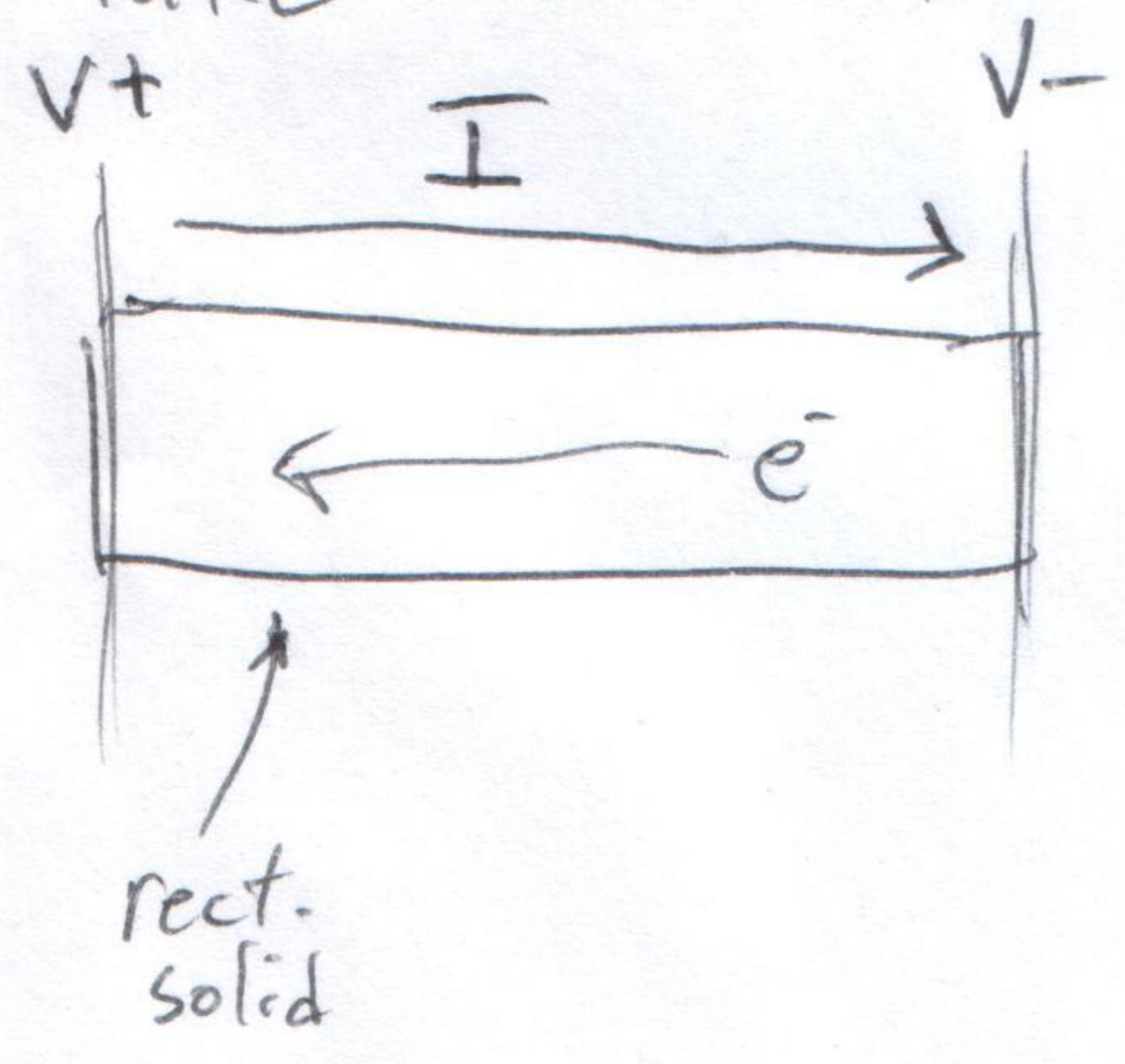
Overview:

- Physics of Conduction
- Quantum Exclusion (IE Band Gap)
- Semiconductors
- Doping
- Junctions
- Diodes
 - Modeling & Context
 - Linear/Nonlinear Combinations
 - Designs
- FETs
 - Modeling & Context
 - Designs
- Logic Design

Labs

- Data Acq./LabView
- Diode Behavior
- Diode Circuit Applications
- Transistor Behavior
- Transistor Circuits
- More Transistor Circuits
- Design Project

So. Let's take a look at resistors.



- motion not ballistic
 - leads to ohm's law
- $$V = IR, R = \sigma \cdot \frac{L}{W \cdot H}$$

Physics of Conduction:

In vacuum:

- throw an electron
- electron goes as $\vec{F} = m\vec{a}$

In nonconductor:

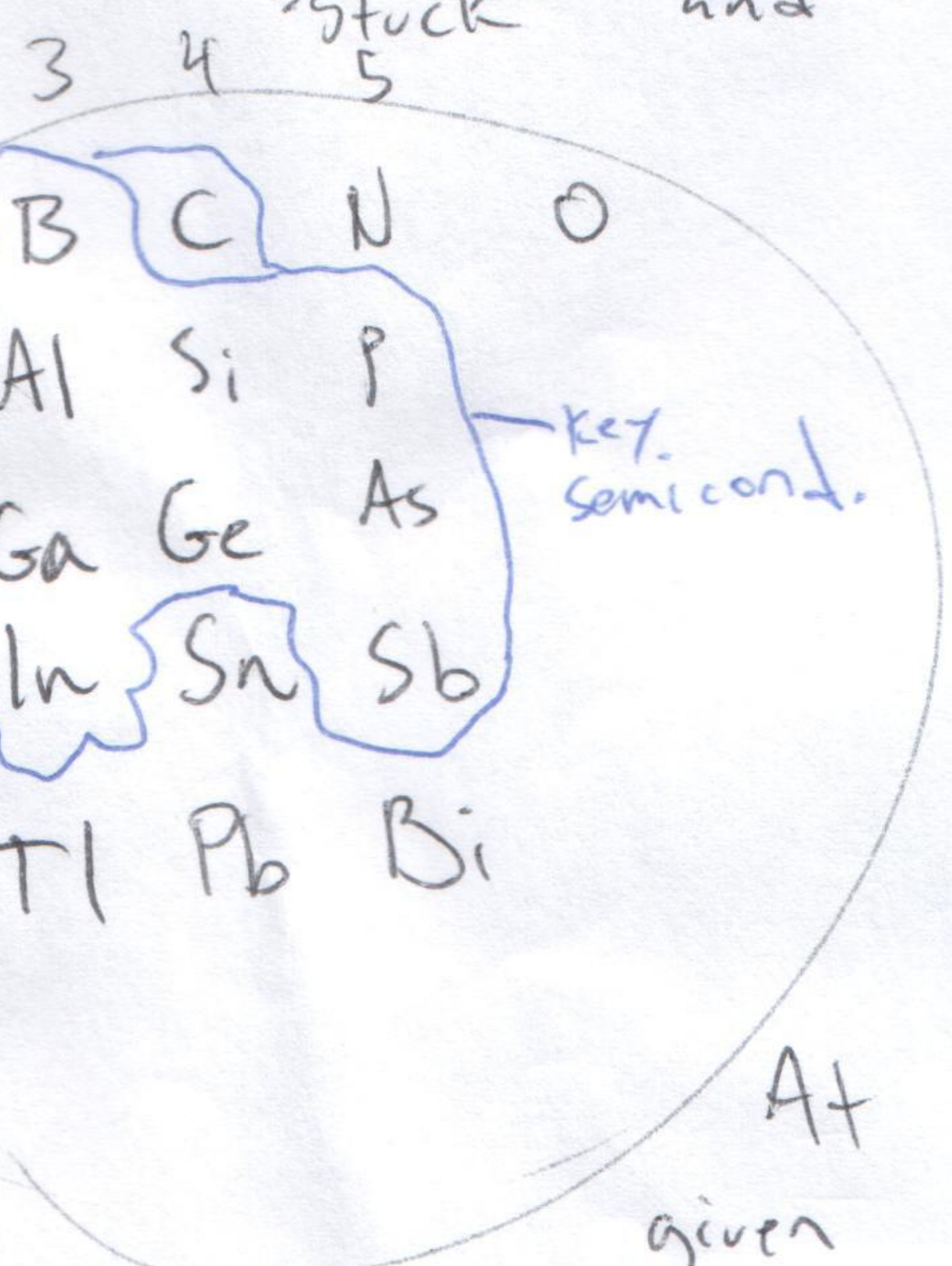
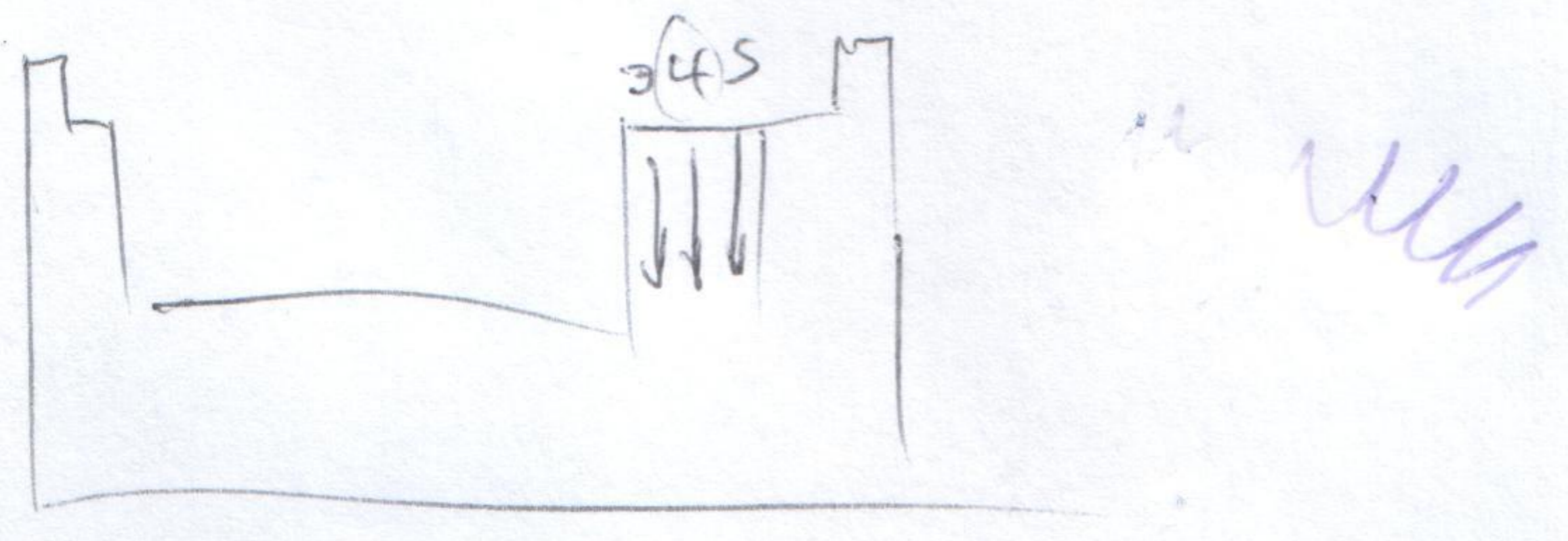
- no available "waves" for electron to join
- "mobility" is low

In metals:

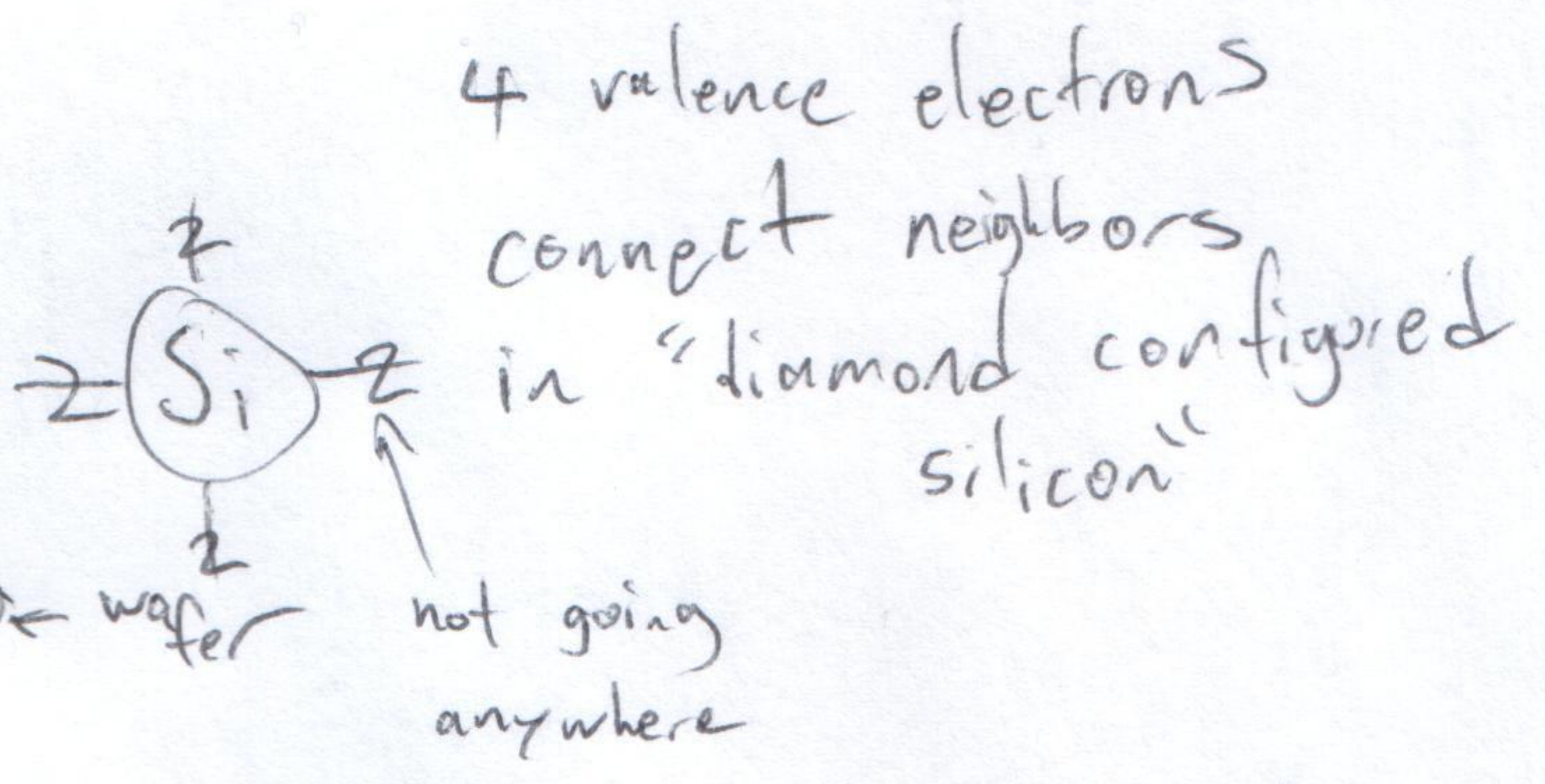
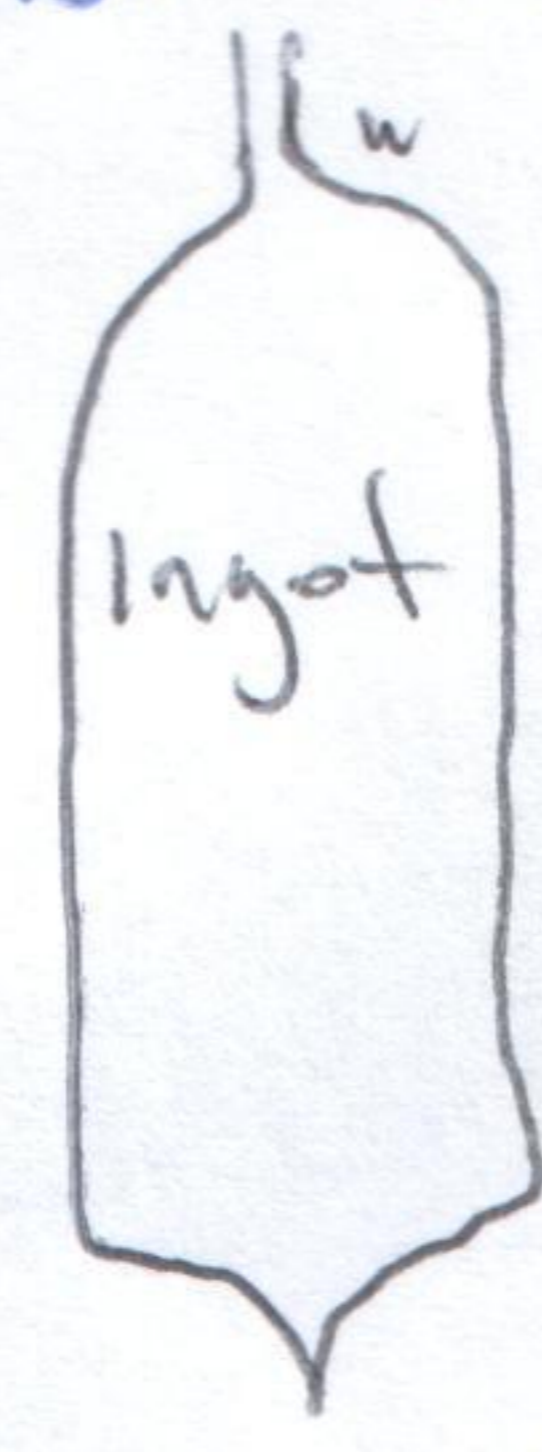
- many orbitals
- electrons very loosely shared
- tiny differences in momentum between "stuck" and "free"

Today:

Resistors, made of semiconductors!



But of course silicon is the big one.



At abs. zero, all vacancies are filled so nothing moves. given any thermal energy, a few will get loose.

Semiconductor Carrier Density as a function of temperature:

$$n_i^2 = BT^3 e^{-\frac{E_g}{kT}} \quad \text{units: } \text{cm}^{-6}$$

material param.
 $1.08 \times 10^{31} \text{ K}^3 \cdot \text{cm}^{-6}$
 in silicon

Boltzmann's constant
 $8.62 \times 10^{-5} \text{ eV/K}$

n_i : # carriers / cm^3

Manipulating p & n: Donors & Acceptors

"Acceptors" $2+$
 B
 Al
 Ga
 In

"Donors" $-$
 N
 P
 As
 Sb

Donors & Acceptors are still electrically neutral but make the lattice adjust to be tilted towards more holes or more ^(free) electrons.

(on their own, these elements tend to form very different crystals than Si, but in mostly-silicon lattices, they fit in.)

$$N_D = \frac{\text{\# atoms}}{\text{cm}^3}$$

$$N_A = \frac{\text{\# atoms}}{\text{cm}^3}$$

To not ionize, we must have:

$$q(N_D + p - N_A - n) = 0$$

At thermal equilibrium, we will also have:

$$pn = n_i^2$$

Negative Doping

< P, As, Sb >
 Phosphorus Arsenic Antimony

$$N_D \gg N_A$$

$$n = \frac{N_D + \sqrt{N_D^2 + 4n_i^2}}{2}, \quad p = \frac{n_i^2}{n}$$

Positive Doping

< B, Al, Ga, In >

$$N_A \gg N_D$$

$$p = \frac{N_A + \sqrt{N_A^2 + 4n_i^2}}{2}, \quad n = \frac{n_i^2}{p}$$

Practically:

n-type:

$$n \approx N_D$$

$$p \approx \frac{n_i^2}{N_D}$$

p-type:

$$p \approx N_A$$

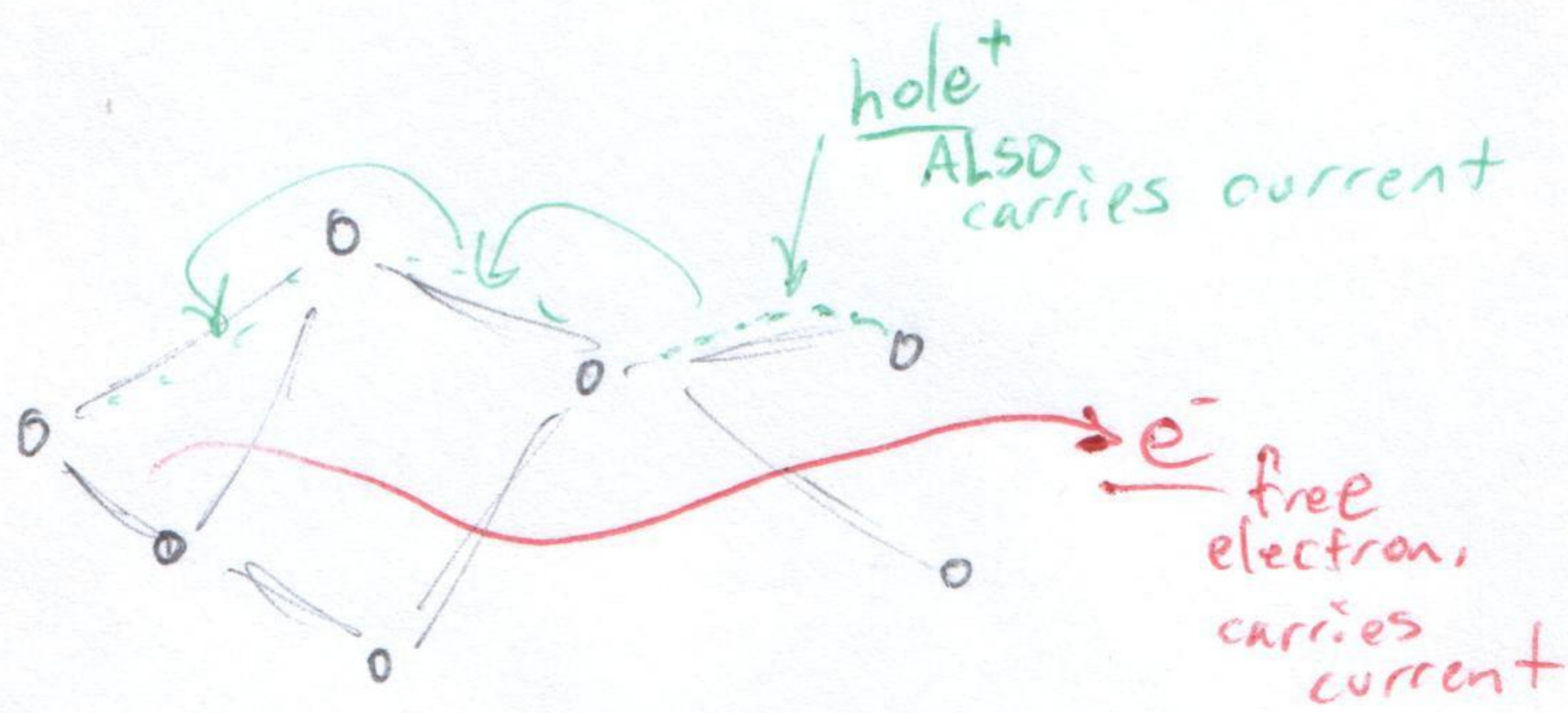
$$n \approx \frac{n_i^2}{N_A}$$

$n_i \approx 10^{10}$, silicon @ room temperature (6.73×10^9)

Holes & Electrons

$$p \cdot n = n_i^2$$

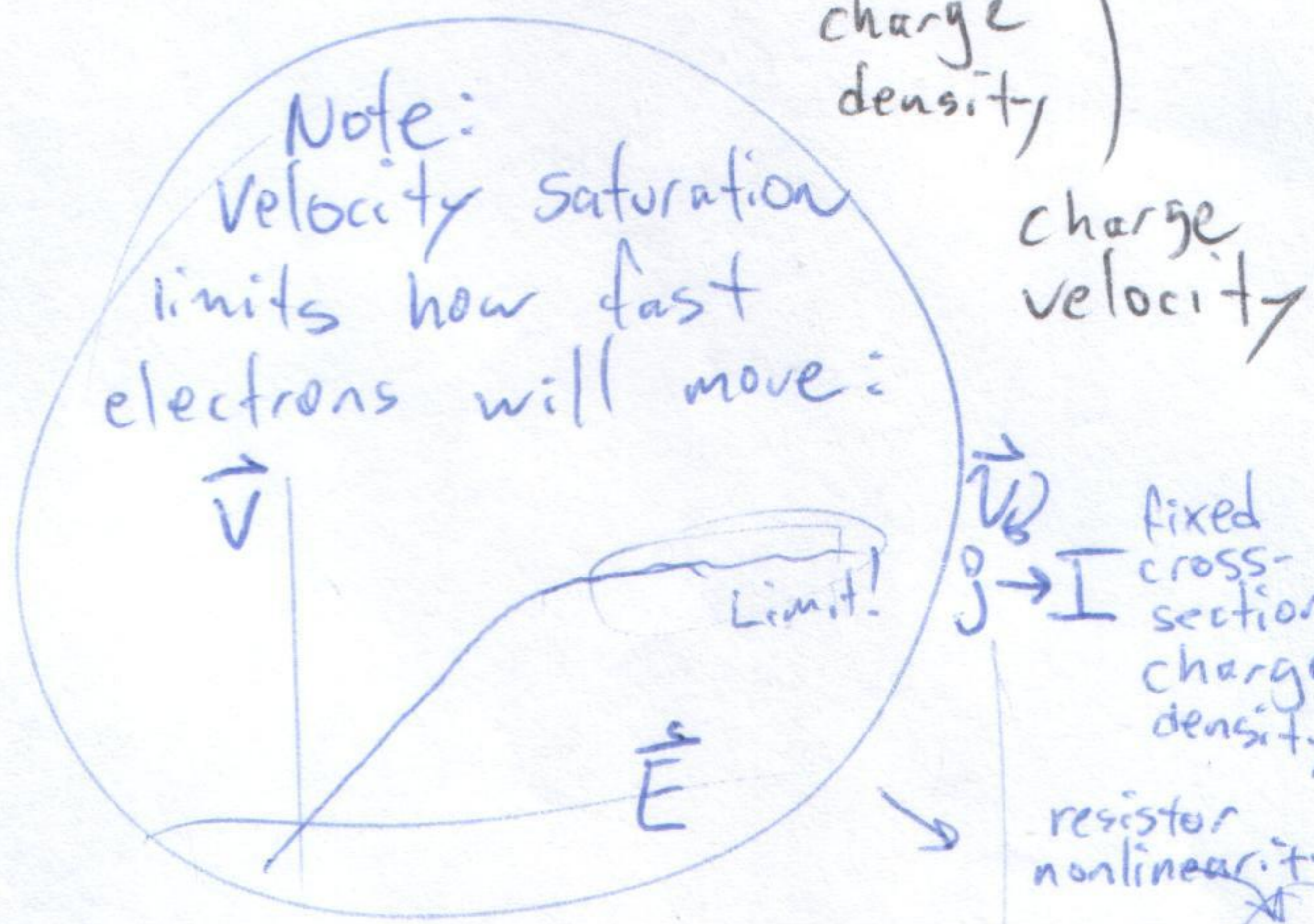
true for all semiconductors in thermal equilibrium.



Drift Current: All currents are Coulomb so current density is $\frac{C}{S \cdot cm^2}$ or $\frac{A}{cm^2} \rightarrow \vec{j} = Q \vec{v}$

Charges in motion for DRIFT:

$$\vec{v}_p = \mu_p \vec{E}, \quad \vec{v}_e = -\mu_e \vec{E}$$

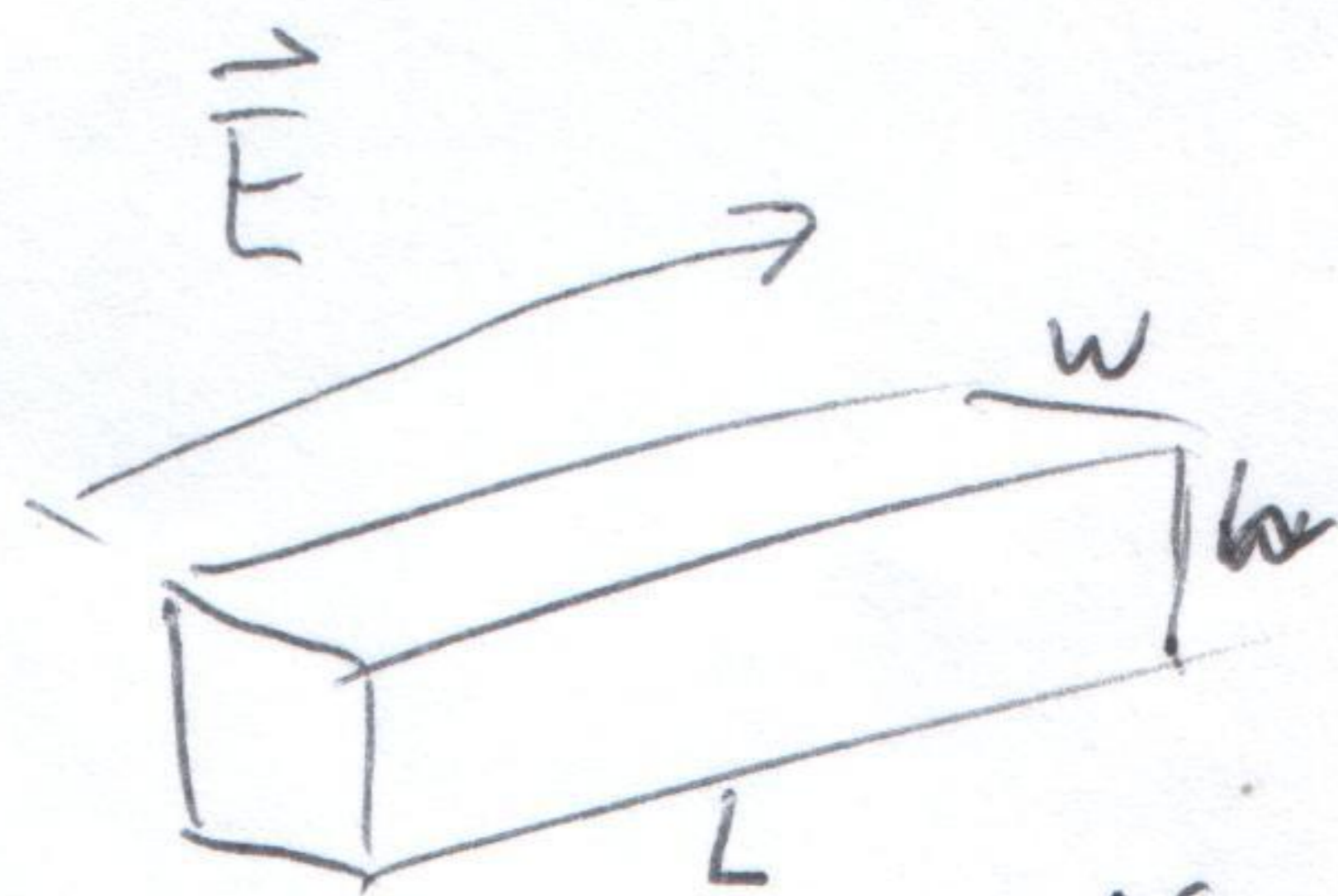


$$j_{drift} = (-qn)(-\mu_n E) + (qP)(\mu_p E)$$

$\leftarrow e^-$ $\xrightarrow{h^+}$
 \xrightarrow{i} \xrightarrow{i}

$$j = q(n\mu_n + p\mu_p)E$$

$\sigma, \Omega/cm$



$$V = L \cdot E$$

$$j = \frac{I}{Area}$$

$$\frac{I}{h \cdot w} = \sigma \cdot \frac{V}{L}$$

$$V = I \frac{1}{\sigma} \frac{L}{h \cdot w}$$

$\rho = \frac{1}{\sigma}, res.$

fixed cross-section charge density
resistor nonlinearity!
fixed distance

R

$$V = IR \dots$$

$$V \rightarrow \vec{E} \cdot D = \frac{\vec{J} \cdot A}{A} \cdot R$$

physics

particle density

$$\star R: \vec{v}_e = -\mu_e \vec{E}, \quad \vec{v}_p = \mu_p \vec{E}$$

$$\vec{j} = (qn) \mu_n \vec{E}$$

$$\vec{j} = q(n\mu_n + p\mu_p) \vec{E}$$

μ : "mobility"
 $\frac{m/s}{V/M} \rightarrow \frac{L}{V \cdot s}$ AKA: $\frac{L}{V \cdot s}$

$$j = \frac{I}{A}$$

$$j = \frac{I}{w \cdot h}$$

$$I = j(w \cdot h)$$

$$I = (w \cdot h) \sigma \cdot E$$

$$I = w \cdot h \cdot \sigma \cdot \frac{V}{L}$$

$$V = I \left(\frac{L}{w \cdot h \cdot \sigma} \right)$$

AKA: "a resistor"

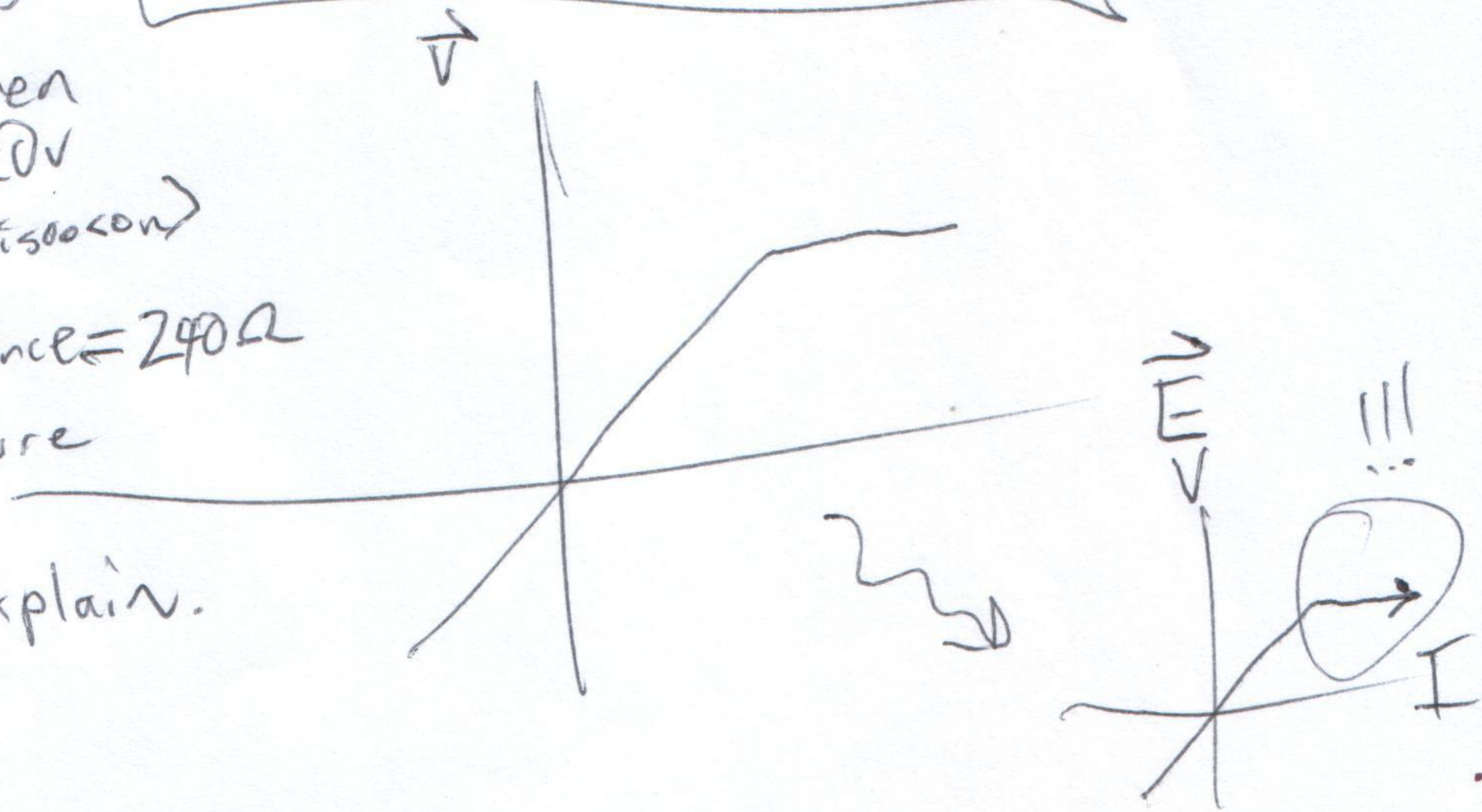
$$V = \int \vec{E}$$

$$V = E \cdot L \quad \xrightarrow{\text{sub } \vec{E} = \frac{V}{L}}$$

Velocity Saturation

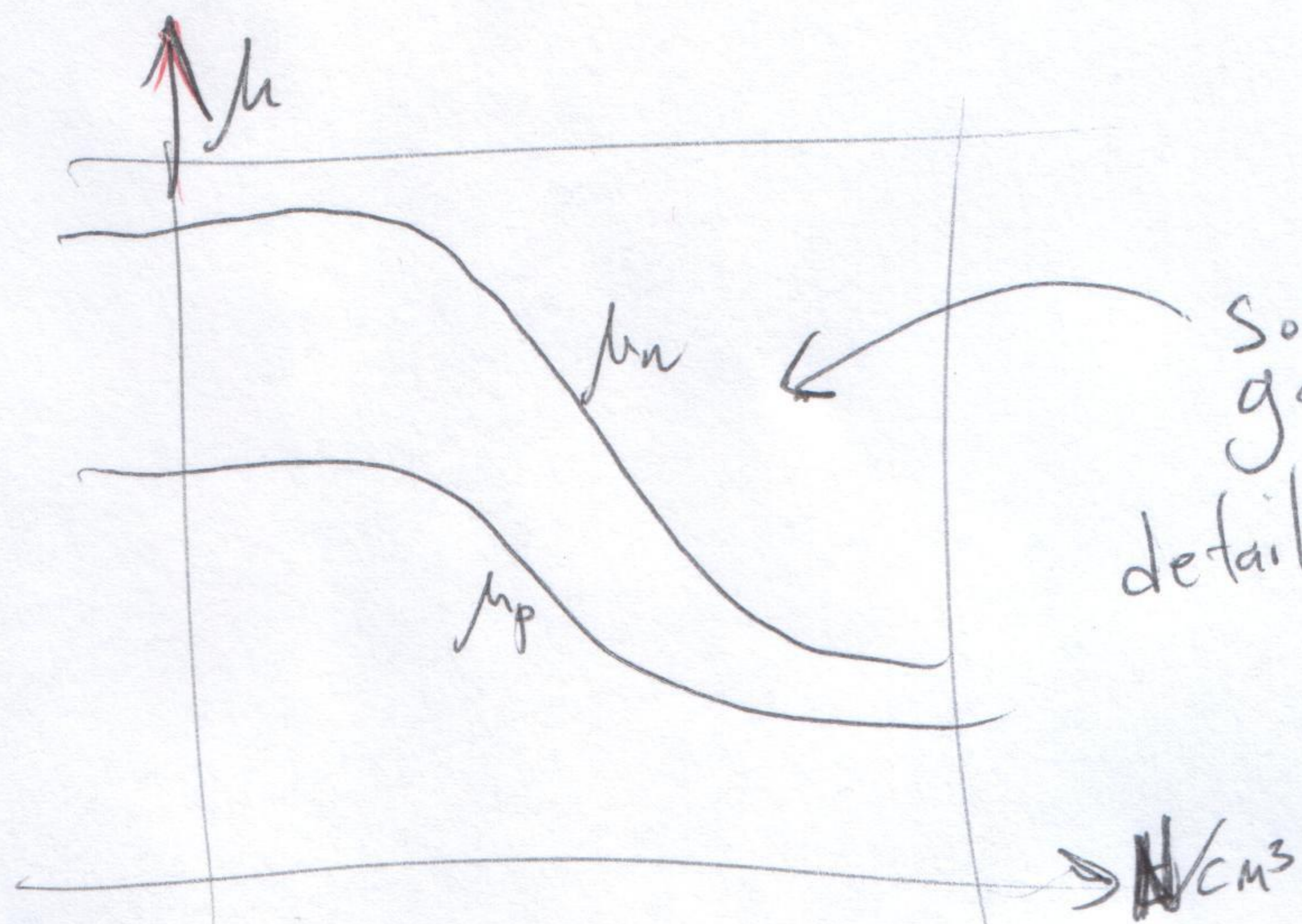
Quiz: if a light bulb runs at 60W when attached to a 120V DC source (Edison con) find its resistance = 240Ω. You unplug and measure with a DMM.

0.01Ω. Explain.



$$\rho = \frac{l}{q(n\mu_n + p\mu_p)}$$

So we also need μ_n, μ_p



Some graph, details? look it up...

Resistance, quick & Dirty: ingot



N type? \longleftrightarrow P type

μ_n, μ_p , look up

$$n = N_D \quad p = \frac{n_i^2}{n}$$

μ_n, μ_p look up

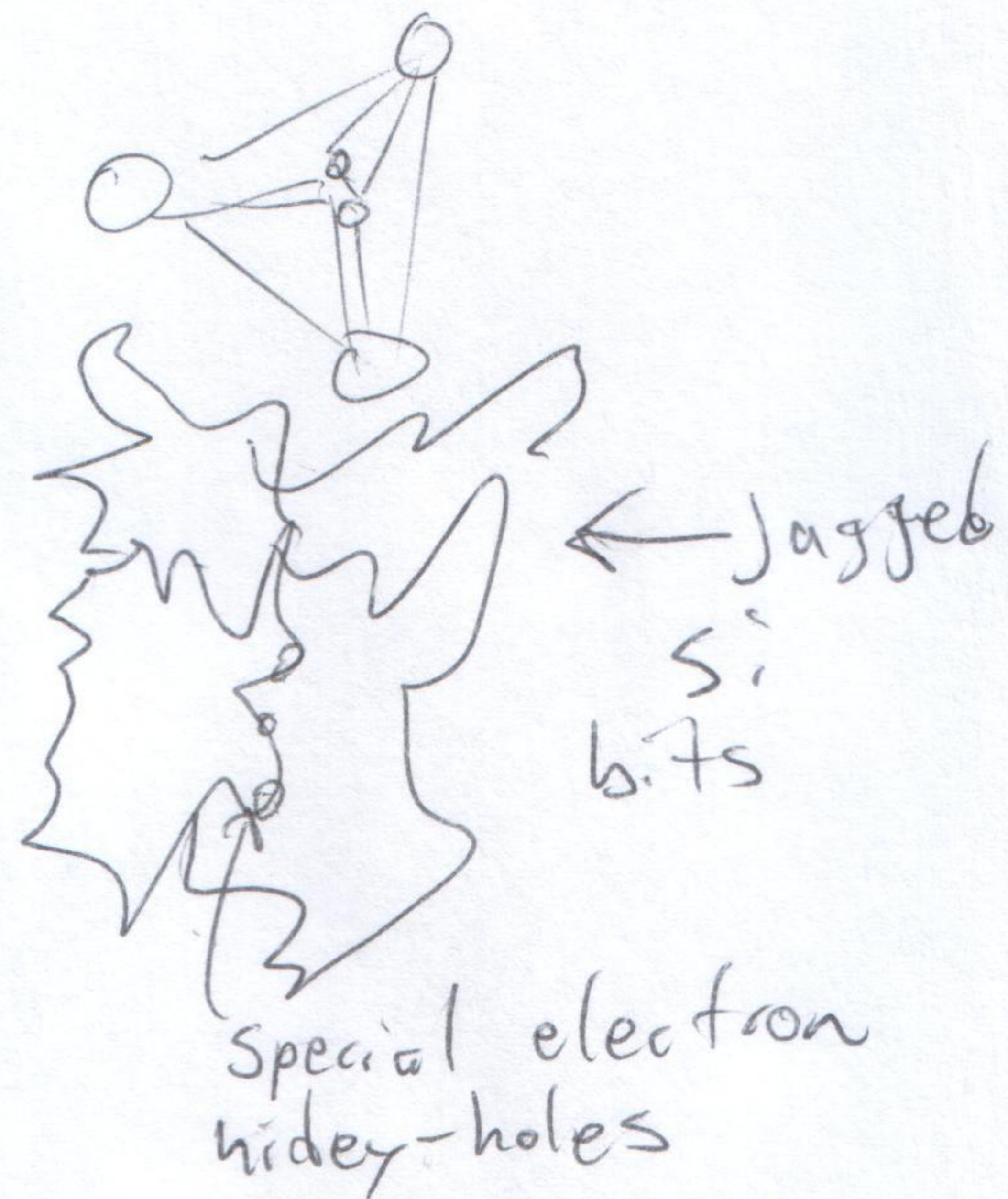
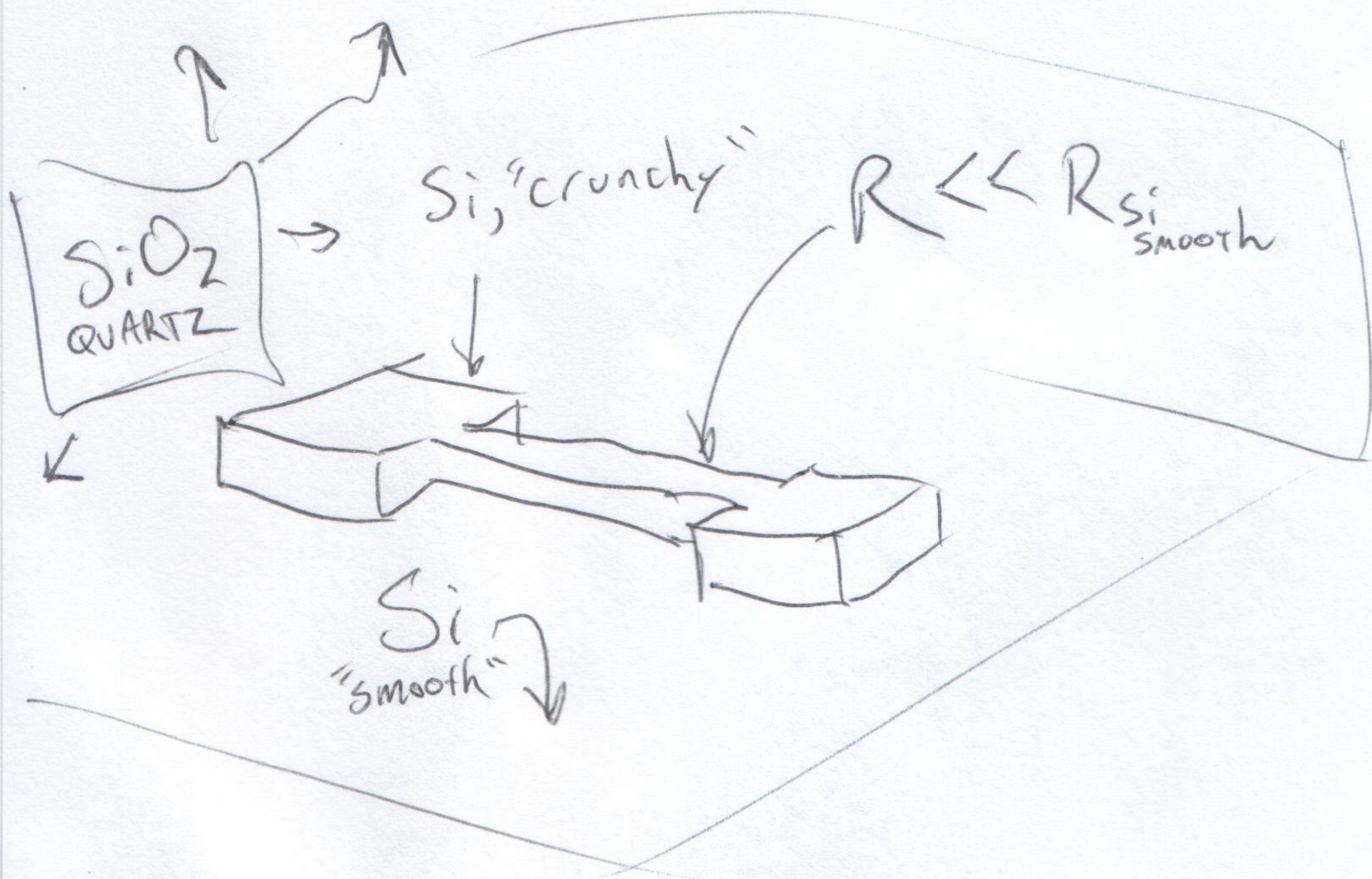
$$p = N_A \quad n = \frac{n_i^2}{p}$$

$$\rho \approx \frac{1}{q\mu_n N_D}$$

$$\rho \approx \frac{1}{q\mu_p N_A}$$

$$\rho = \frac{l}{q(n\mu_n + p\mu_p)} \rightarrow R = \rho \frac{L}{w \cdot H}$$

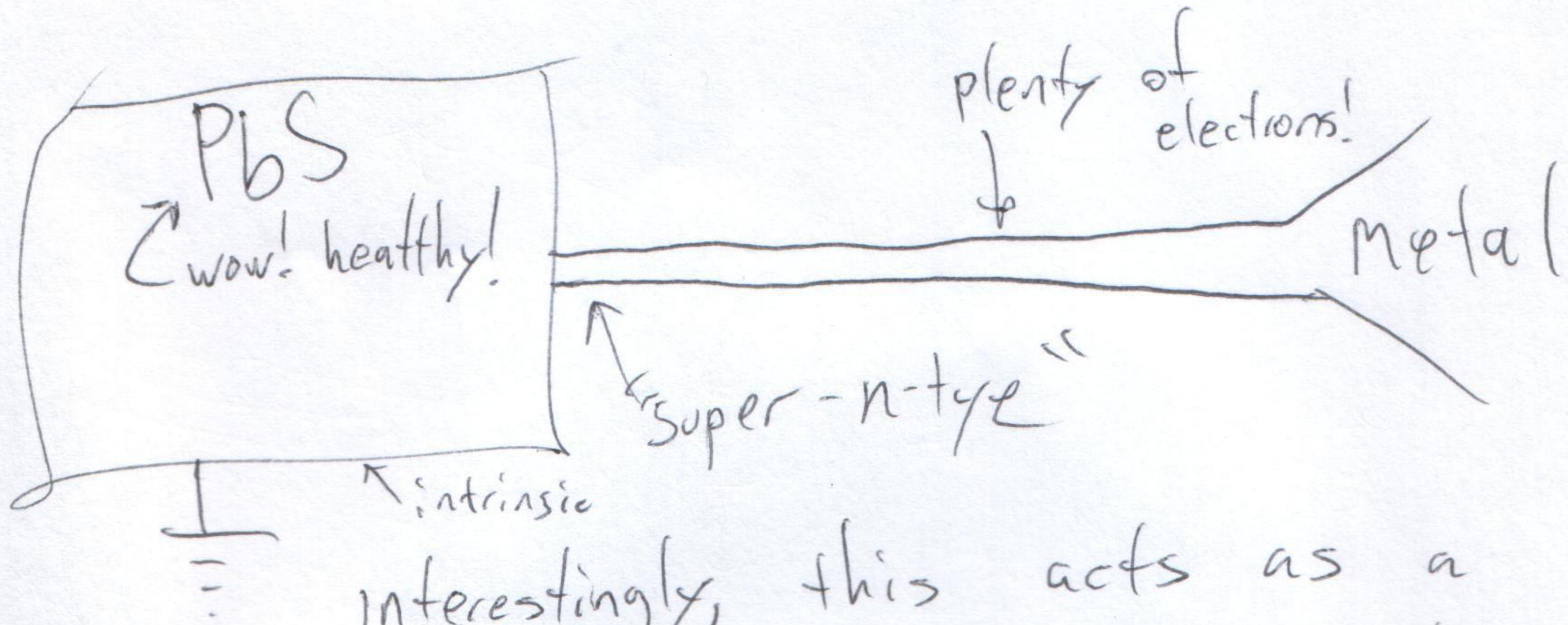
Aside: Polysilicon Resistors



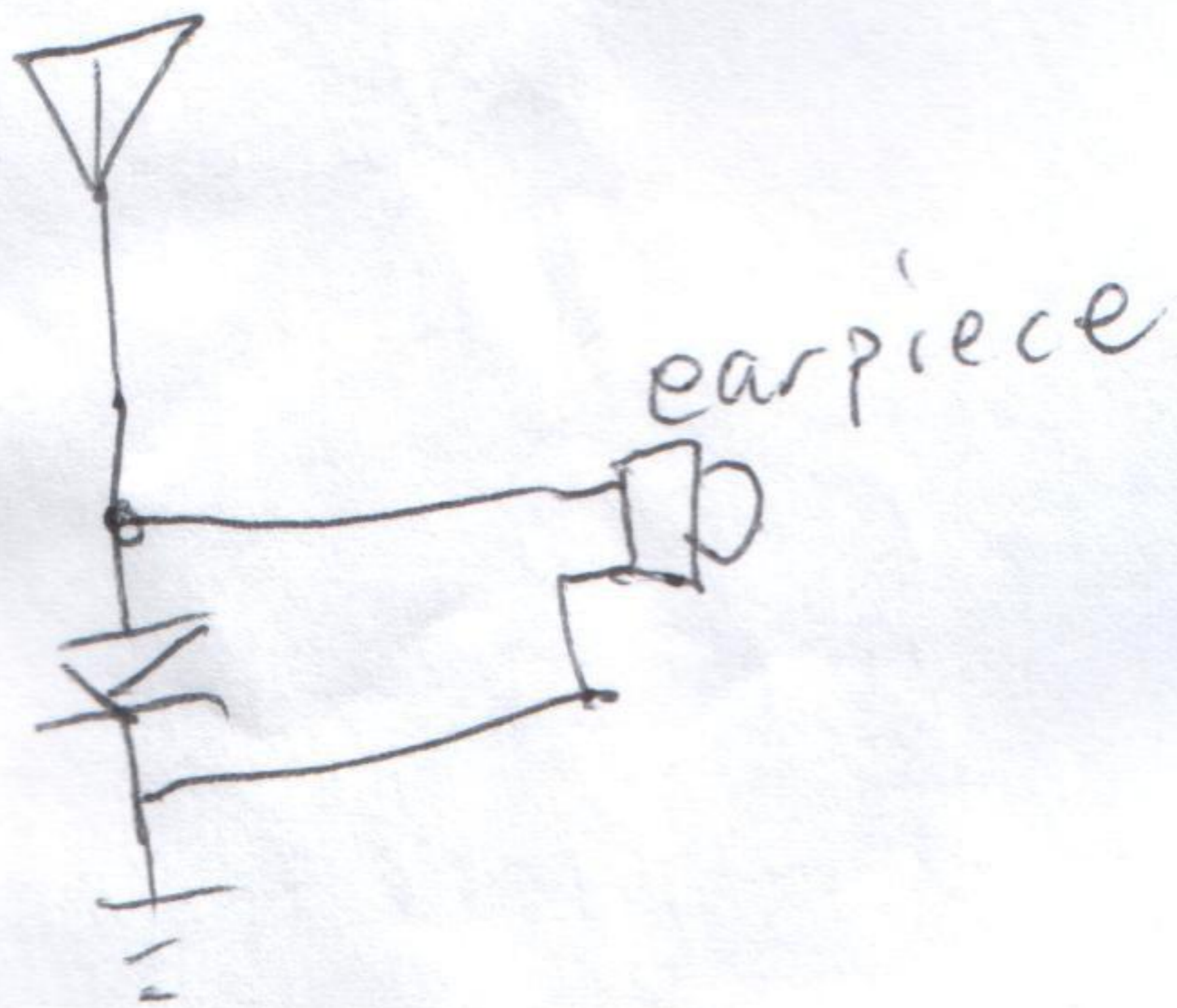
Note: 2D-Fabrication processes
have been LAW for like ~~10~~ years

And then, diodes happened.

Sidebar: the galena diode



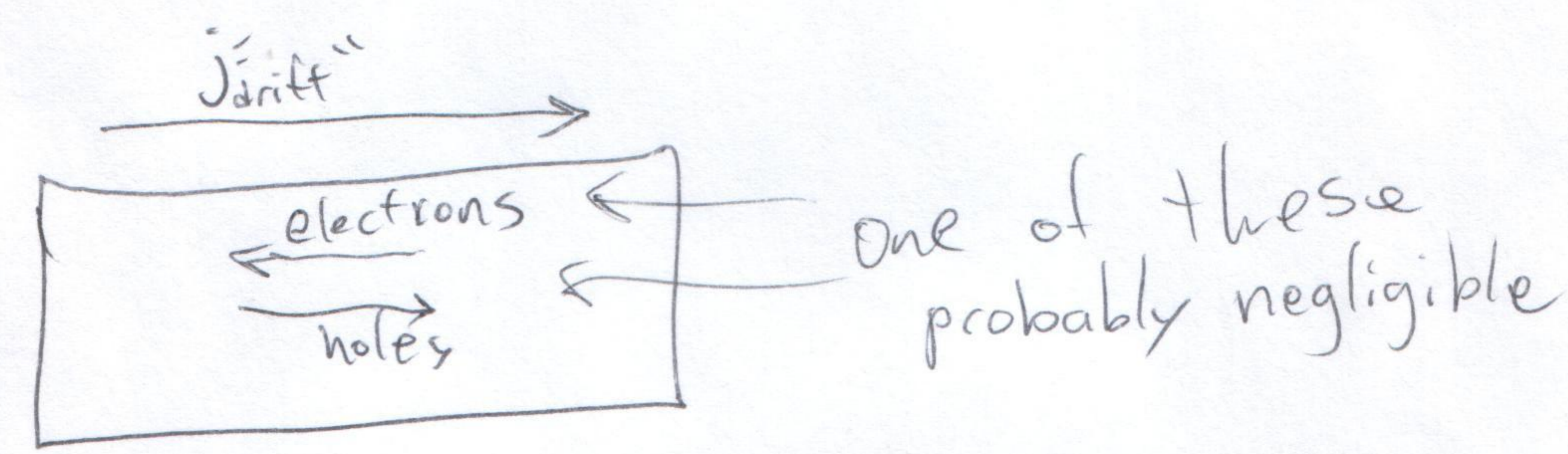
interestingly, this acts as a one way-ish circuit element.



Radio receivers can and have been built in the past by literally attaching quartz and galena to a lot of wire.

This is about as close to minecraft-punk as EE gets.

Quick review of resistance in ^{doped} semicond:



$$j = q \cdot \# \cdot \mu \cdot E \leftarrow \text{e field}$$

\uparrow charge \uparrow population \uparrow mobility, speed/E trade

Resistivity:

$$\sigma \approx \frac{q n \mu_n}{N\text{-type}} \quad \sigma \approx \frac{q N_A \mu_p}{P\text{-Type}}$$

Solids, semiconductors especially, are basically pain-sticks for electrons/holes to "fall" through under an electric field.

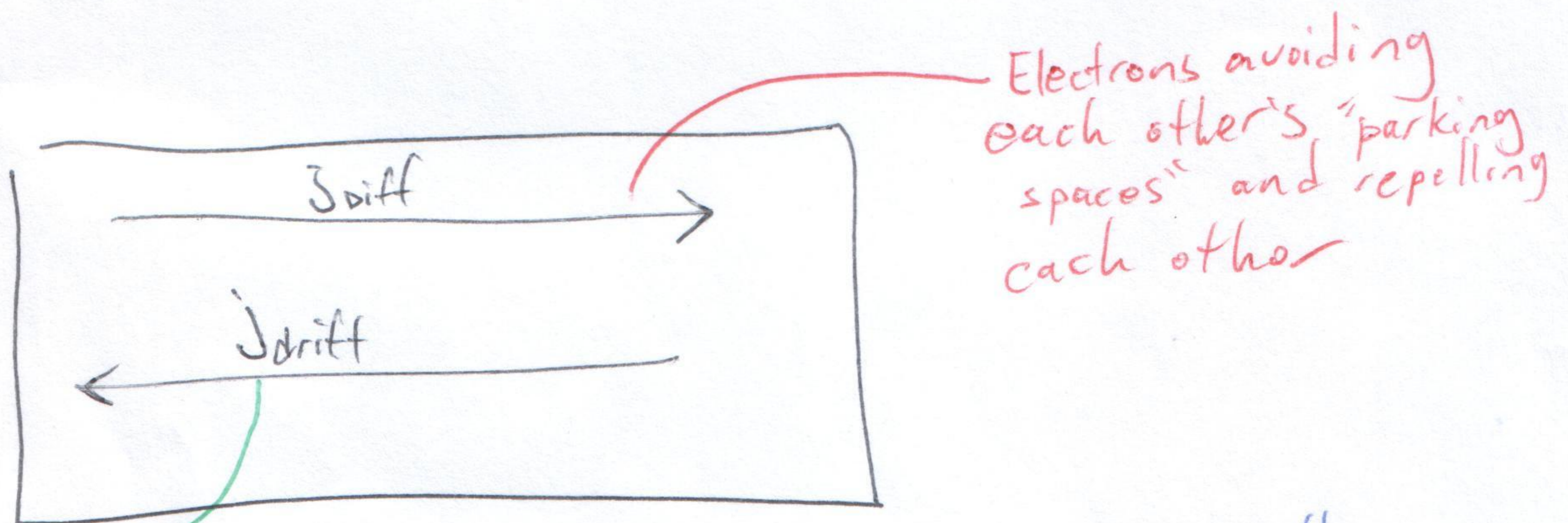
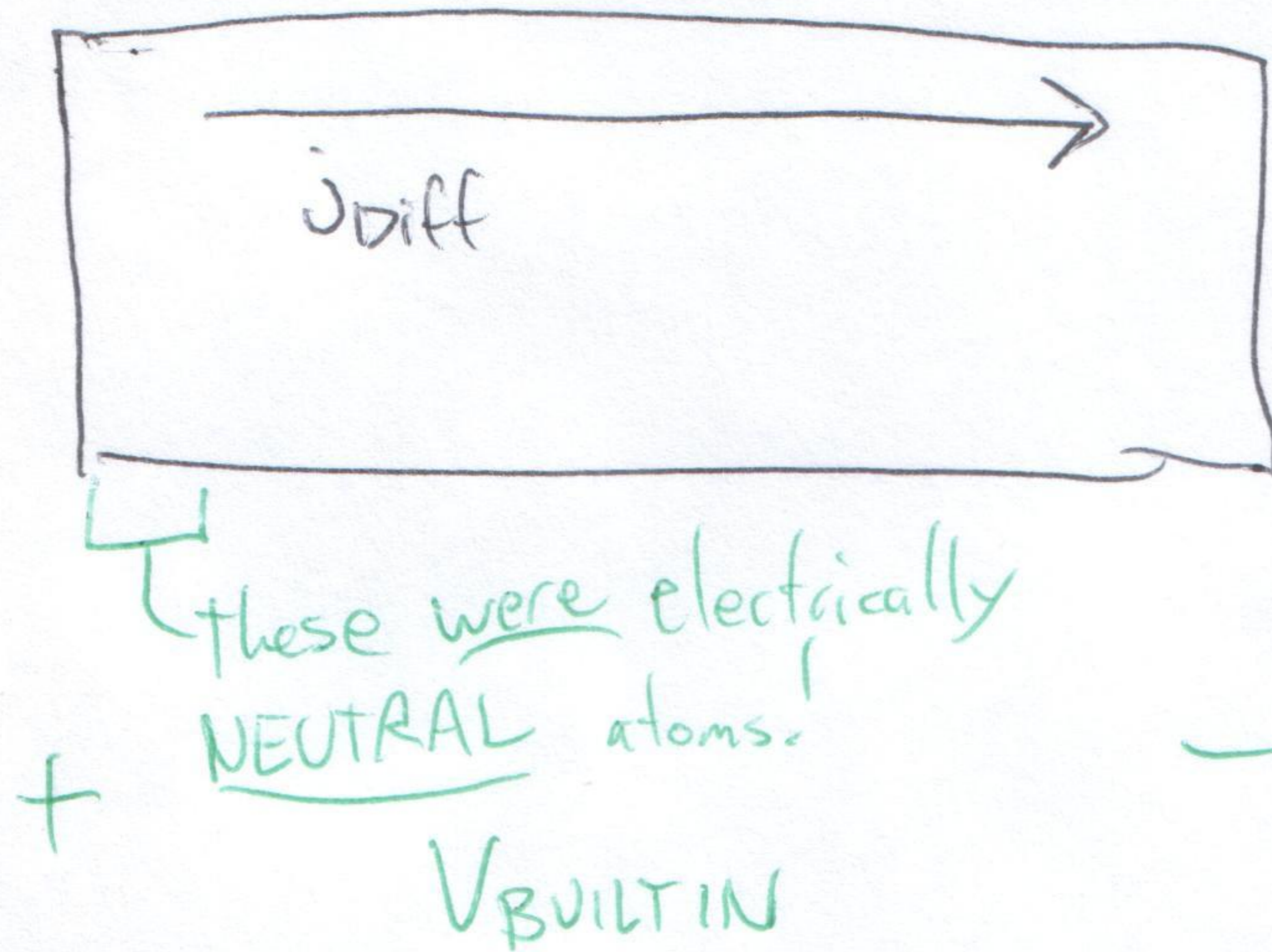
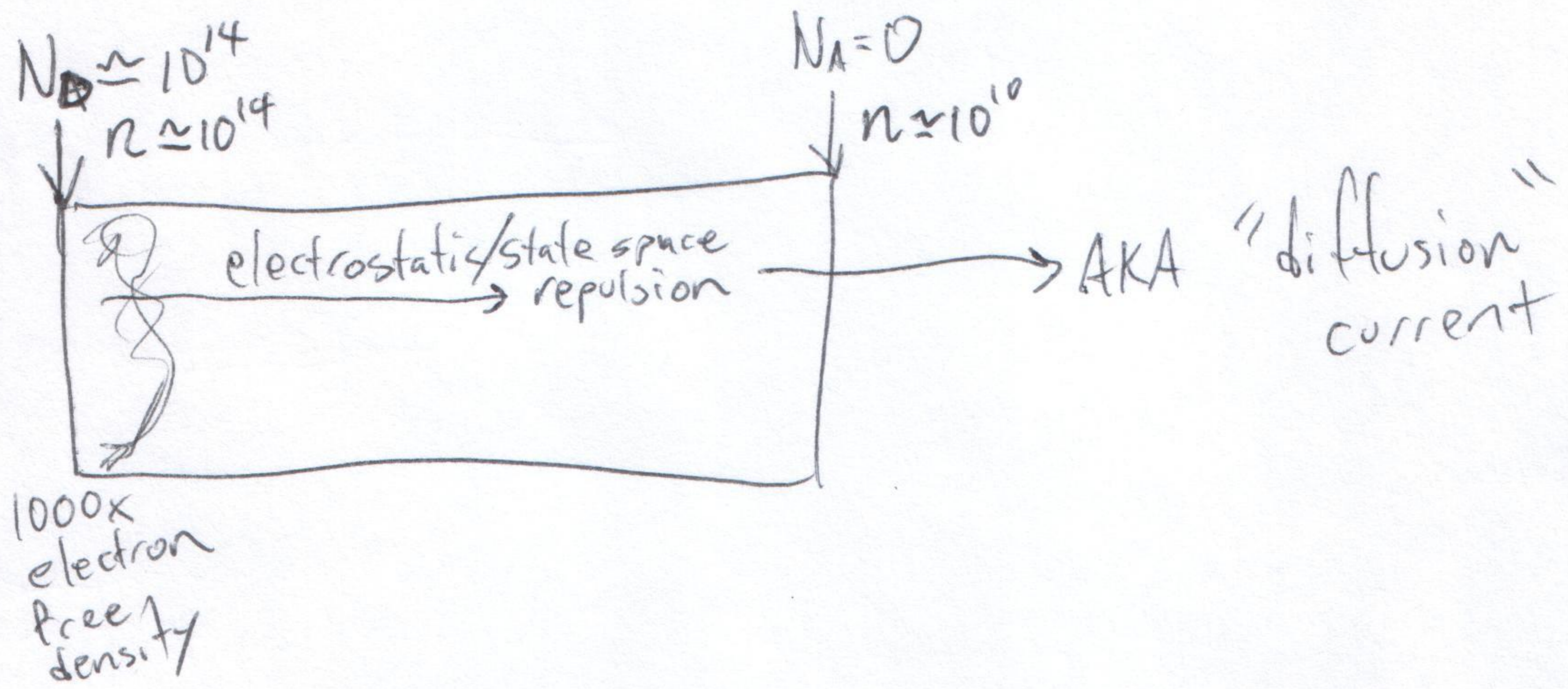
$$p \cdot n = n_i^2$$

Number of charge carriers is $p + n$ so $p = n$ is actually the **LOWEST** number of carriers!

$10^{15} \cdot 10^5 = 10^{20}$	$10^{15} + 10^5 \approx 10^{15}$
$10^5 \cdot 10^{15} = 10^{20}$	$10^5 + 10^{15} \approx 10^{15}$
$10^{10} \cdot 10^{10} = 10^{20}$	$10^{10} + 10^{10} \approx 10^{10}$

Imbalance in carriers makes for more carriers!

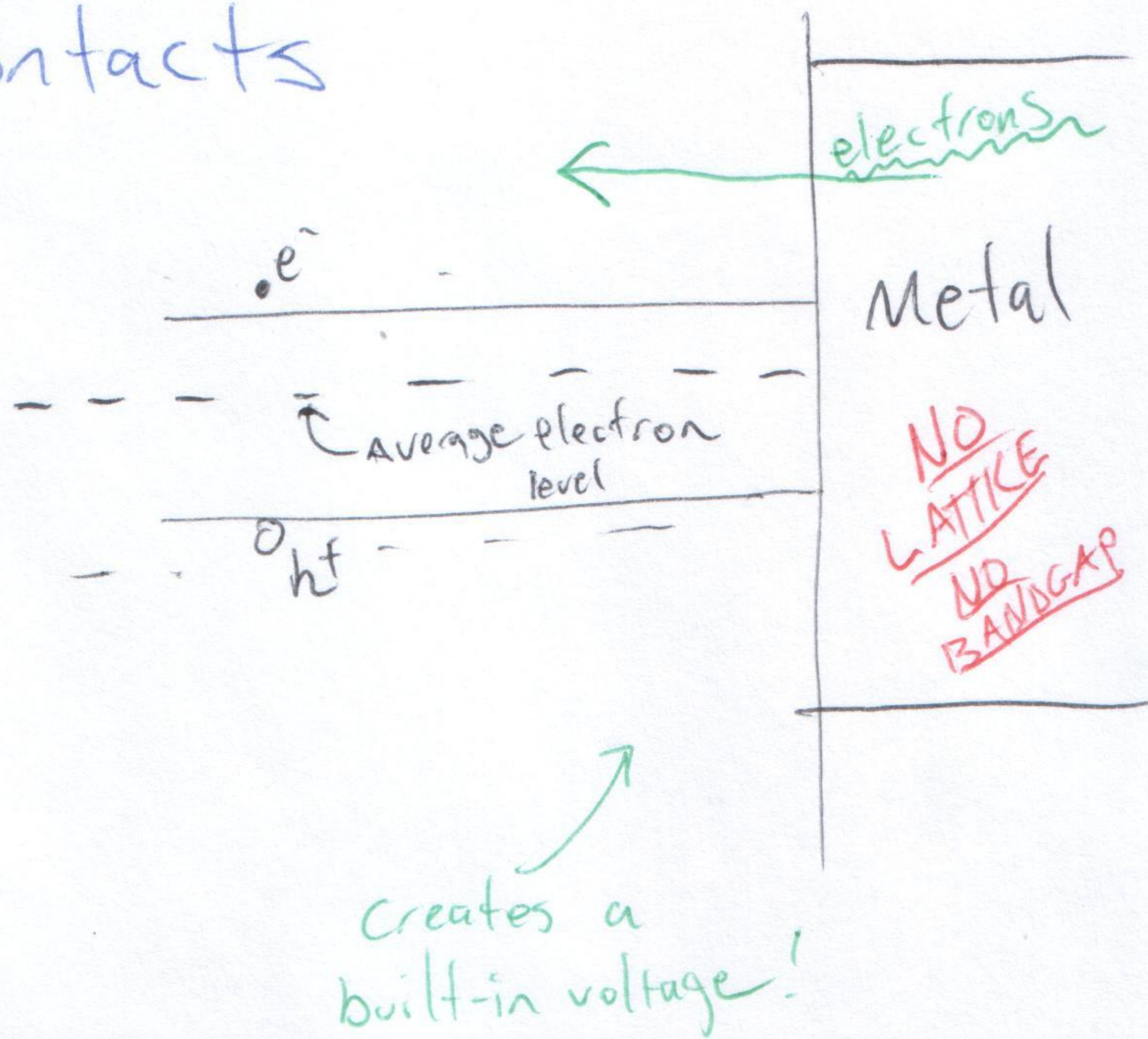
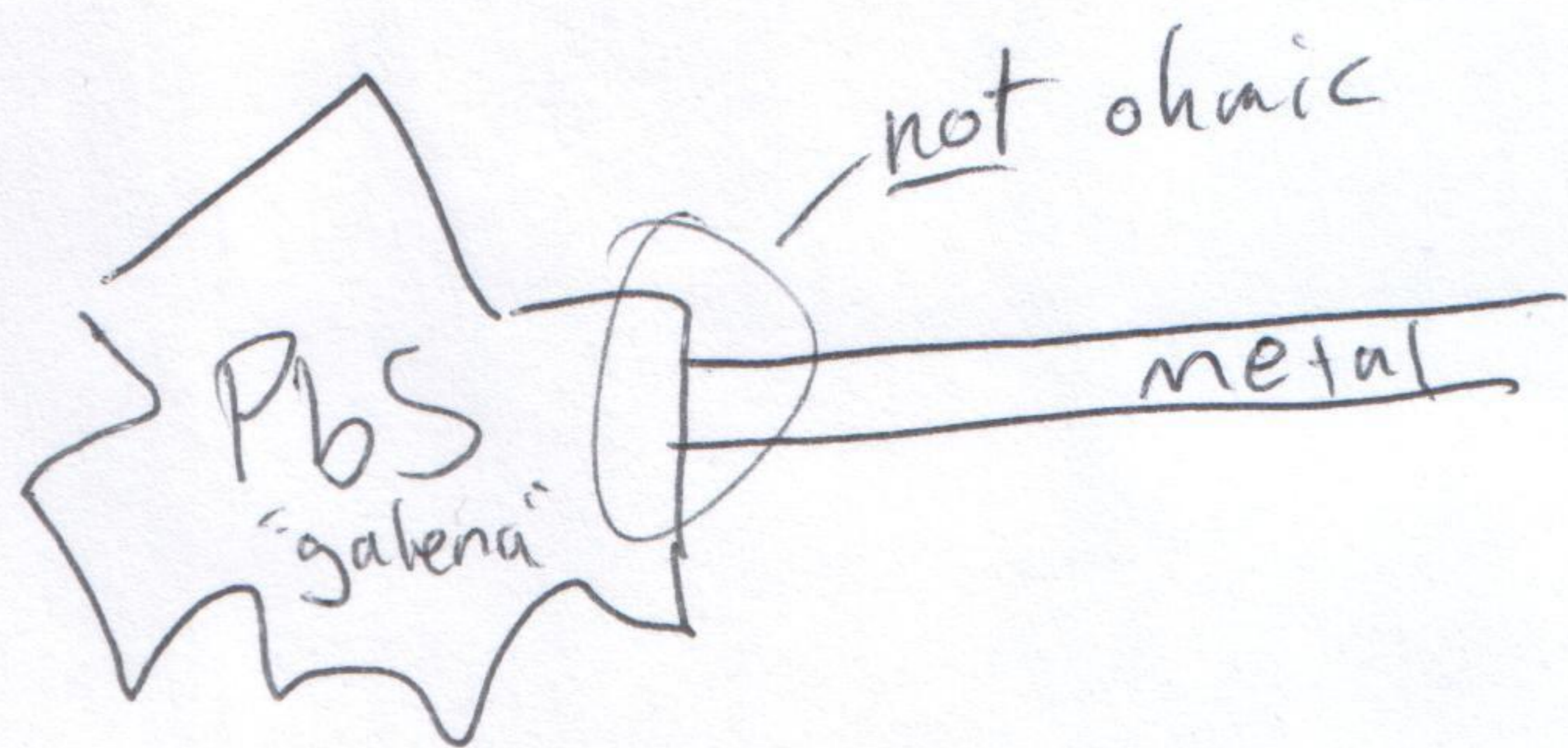
Drift, Diffusion & A block of silicon that thinks it's some kind of battery.



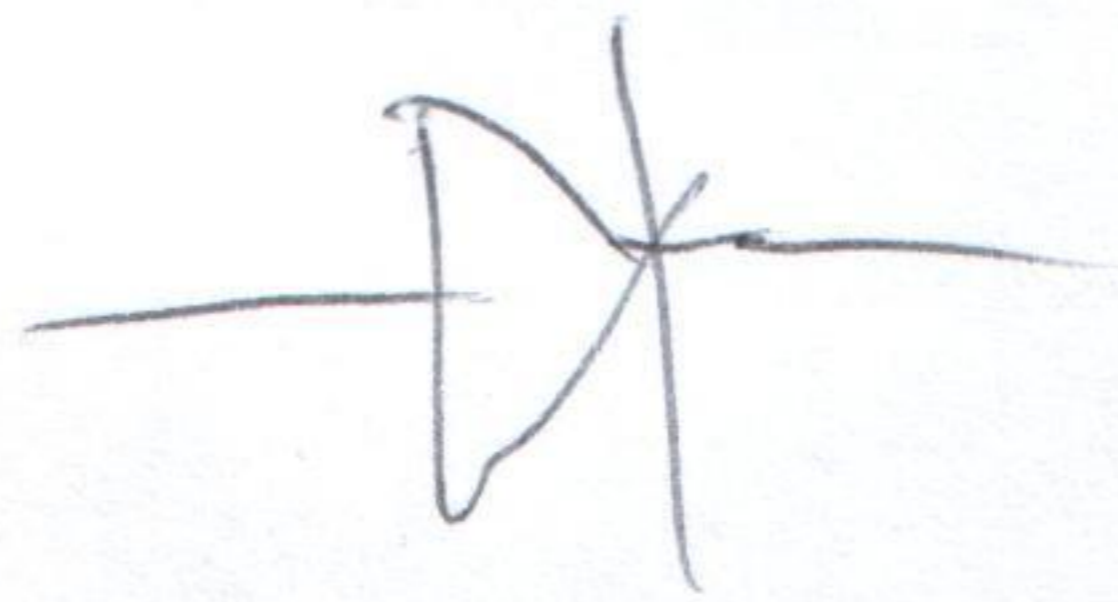
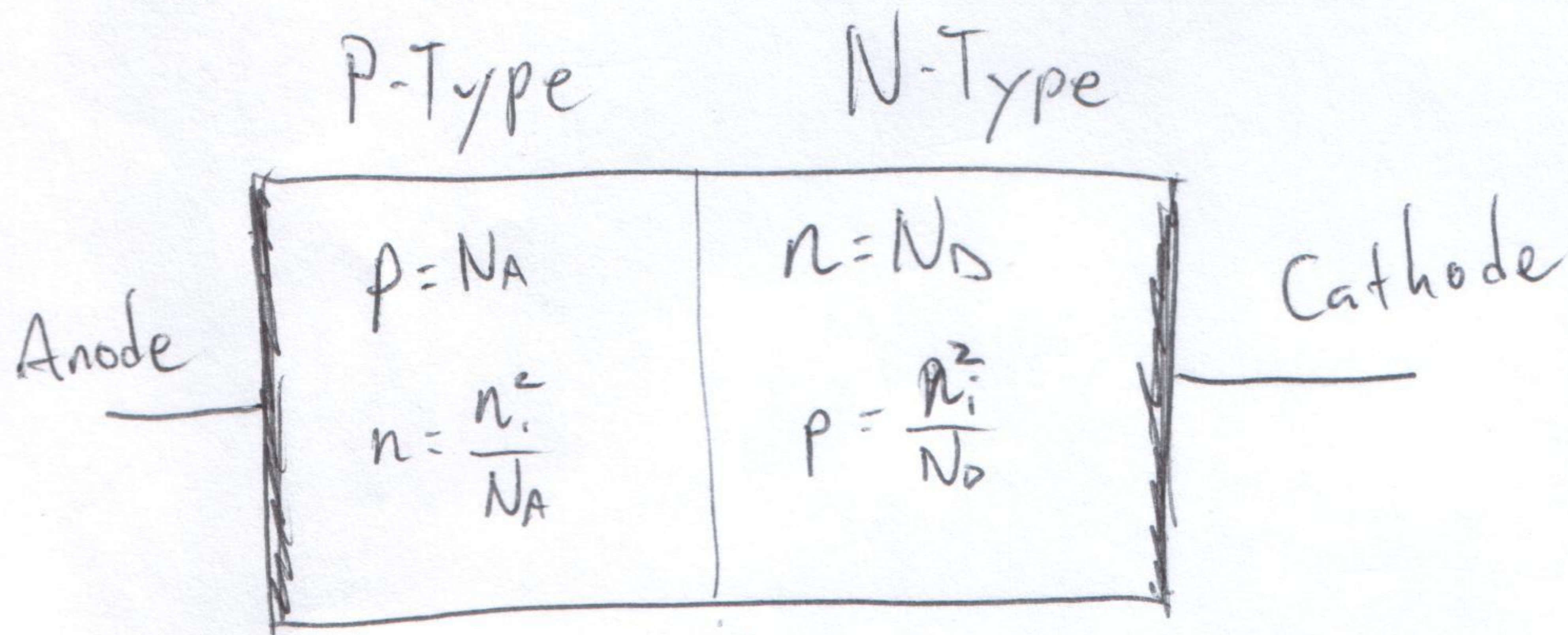
electrons trying to neutralize dopant atoms

@ Equilibrium, no overall current can be present, but voltage can!

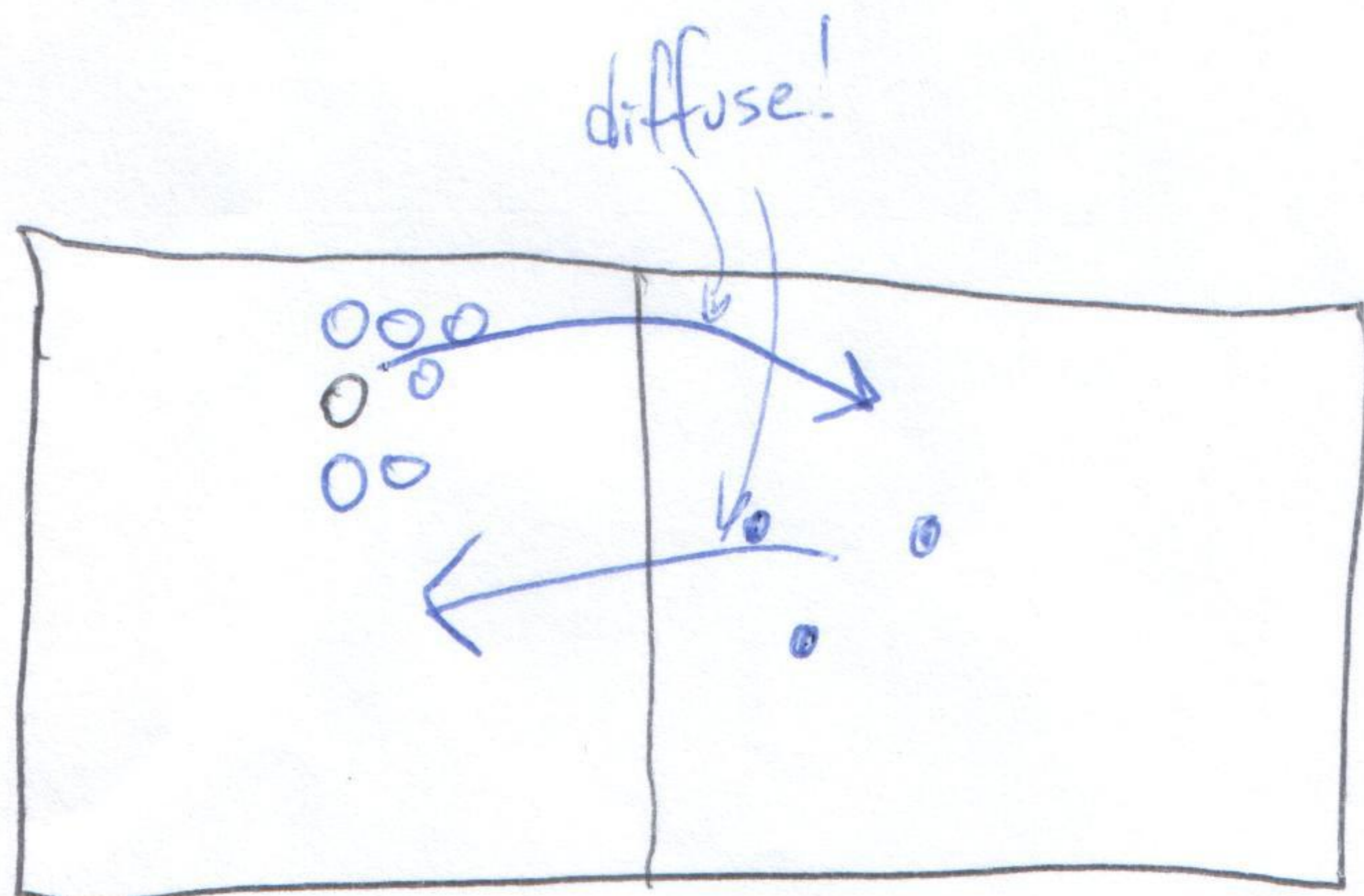
Aside Note: "Ohmic" Contacts

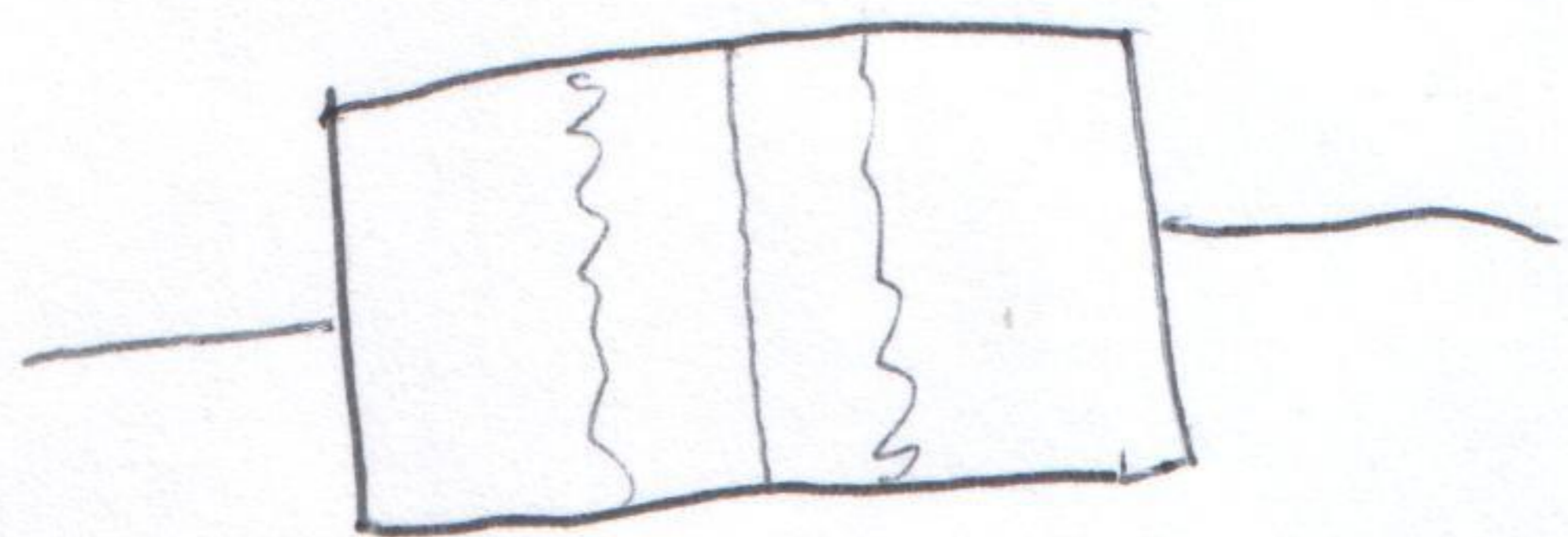


Ohmic Contacts must be designed into a process by tweaking the doping levels. (we will assume this was done.)

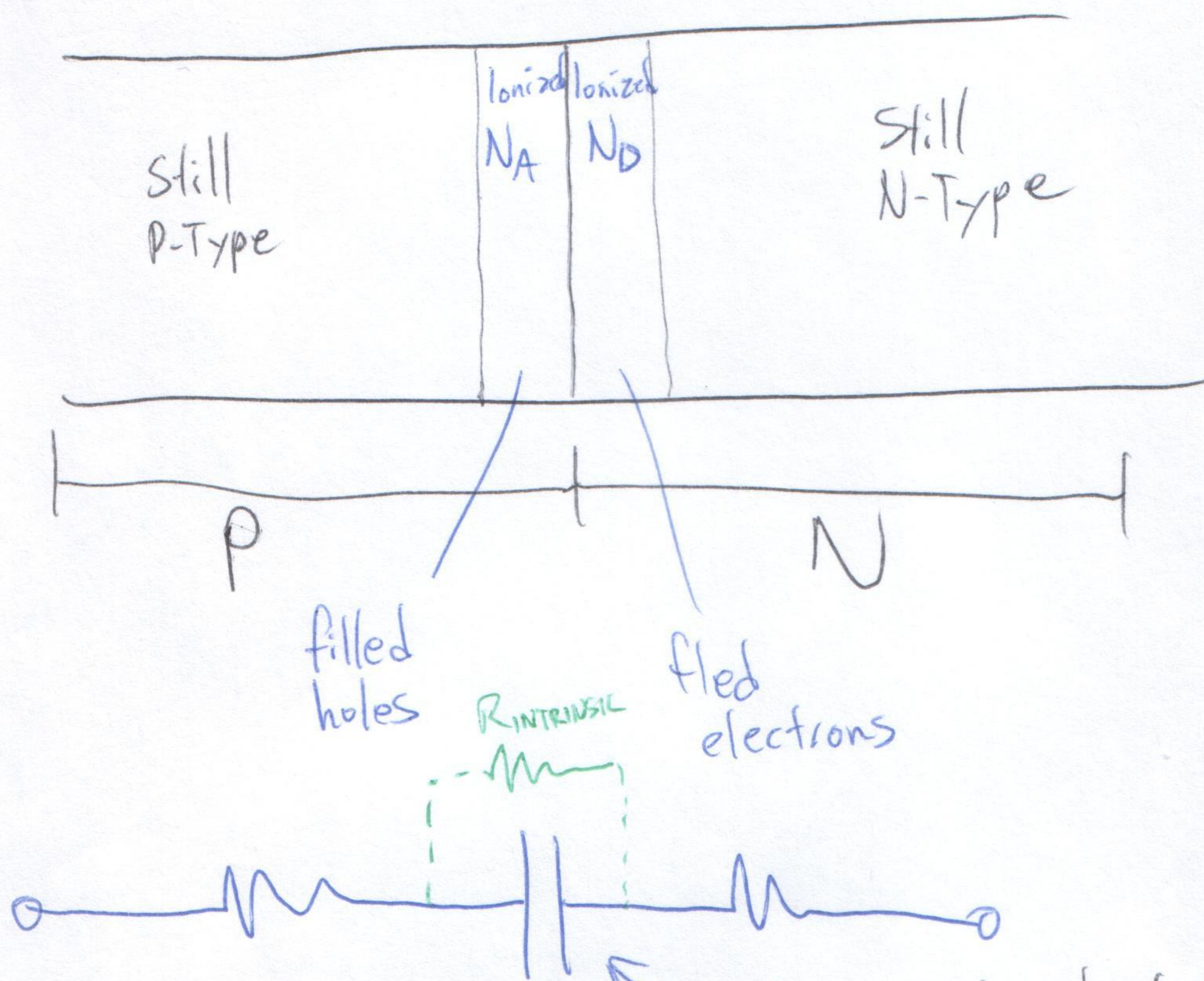


... ever see "Braveheart"?



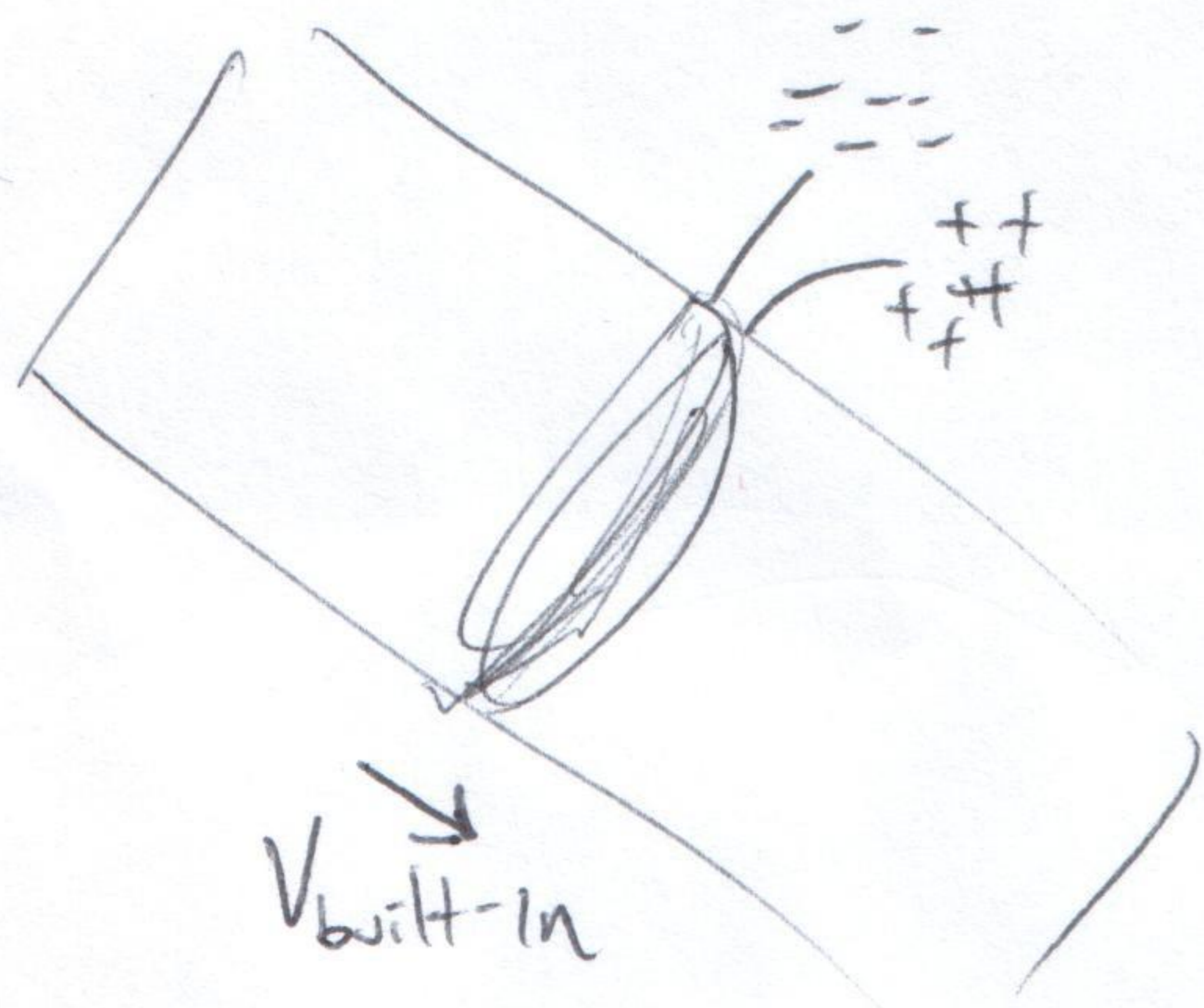


when the dust settles, electrons & holes have canceled each other out:

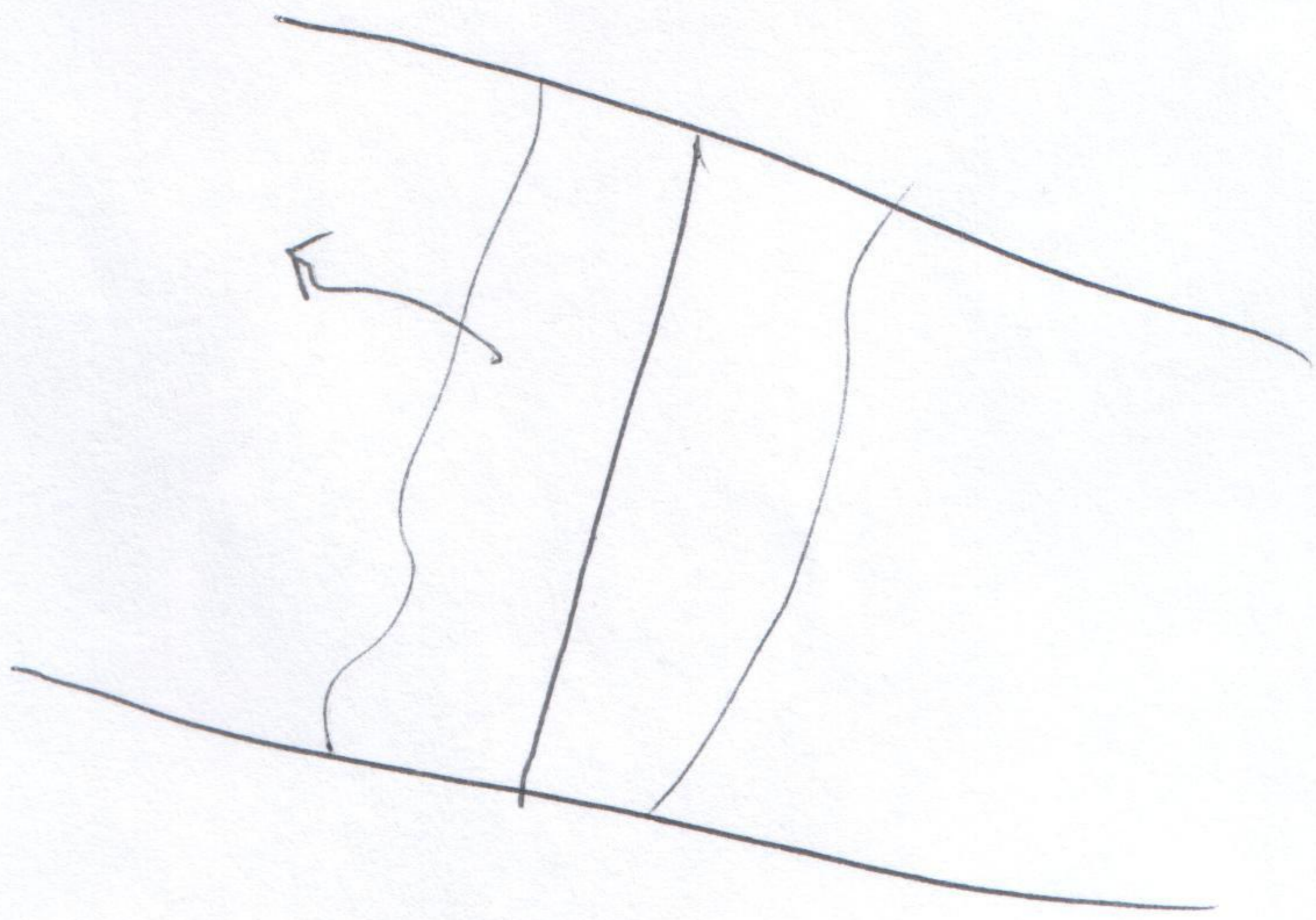
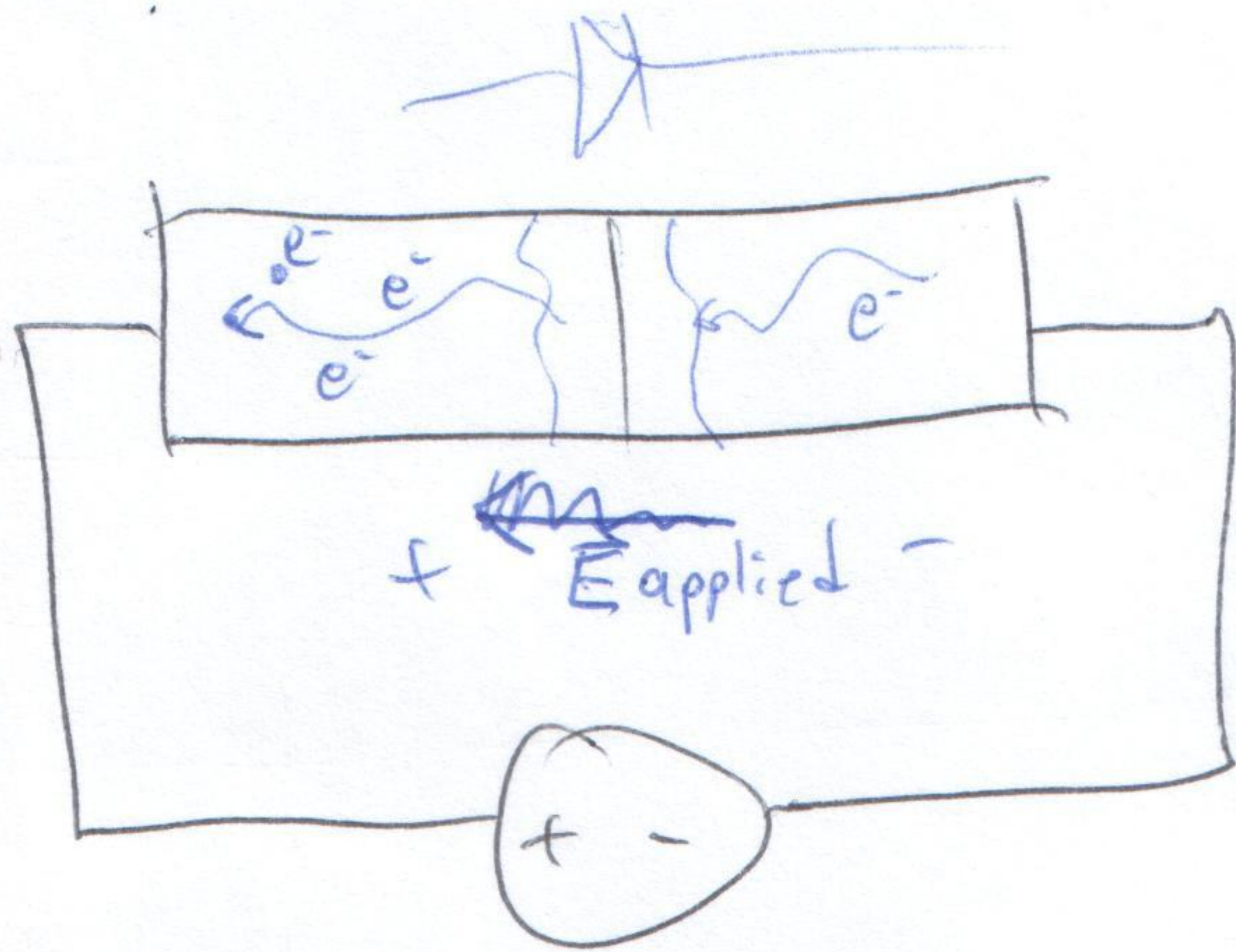
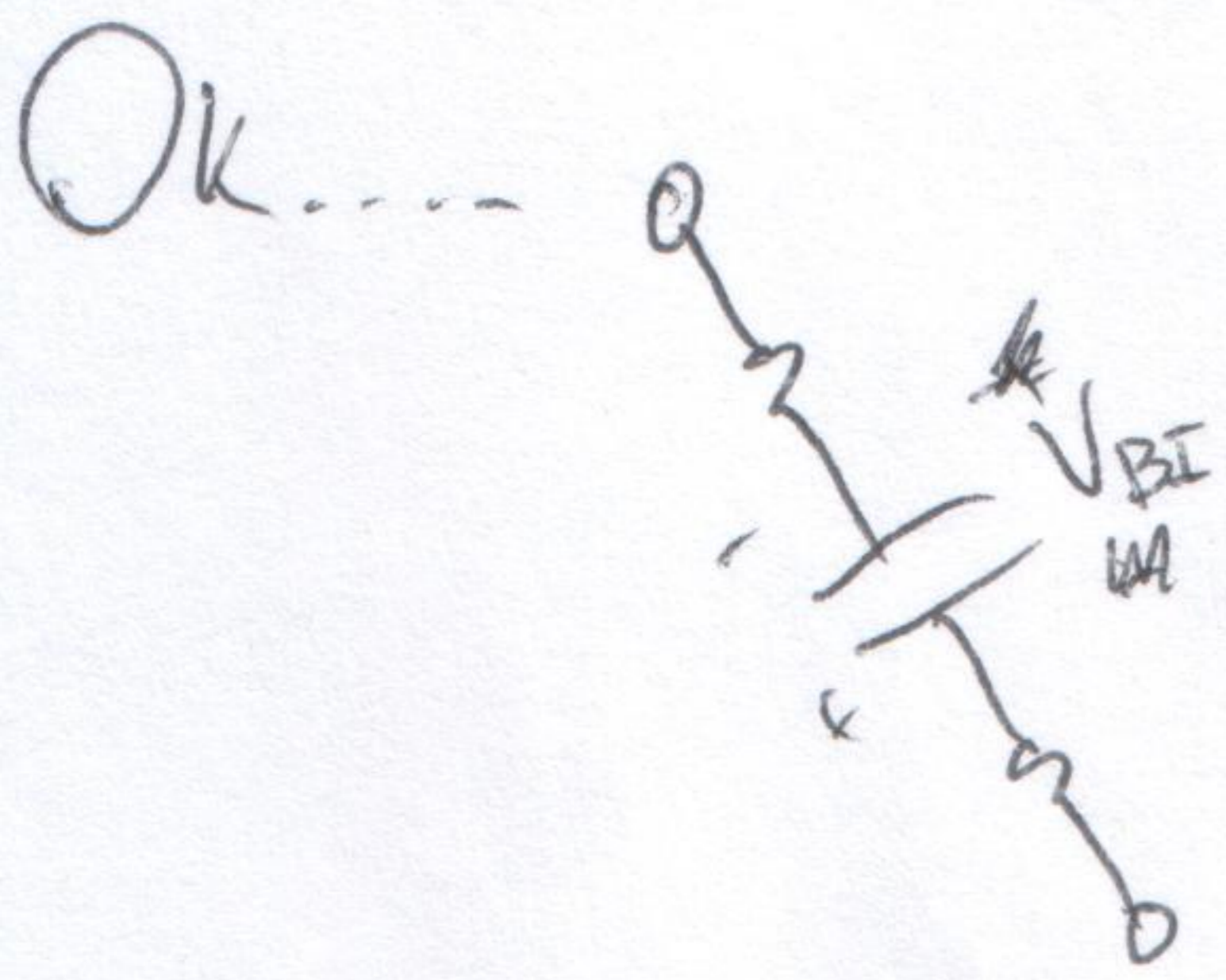


This semiconductor junction has just SPROUTED A CAPACITOR

So that's weird, let's make it worse:



Just like the unevenly-doped slab, this has a built-in voltage, analogous to a charged capacitor, but more COMPLEX than that because charge.

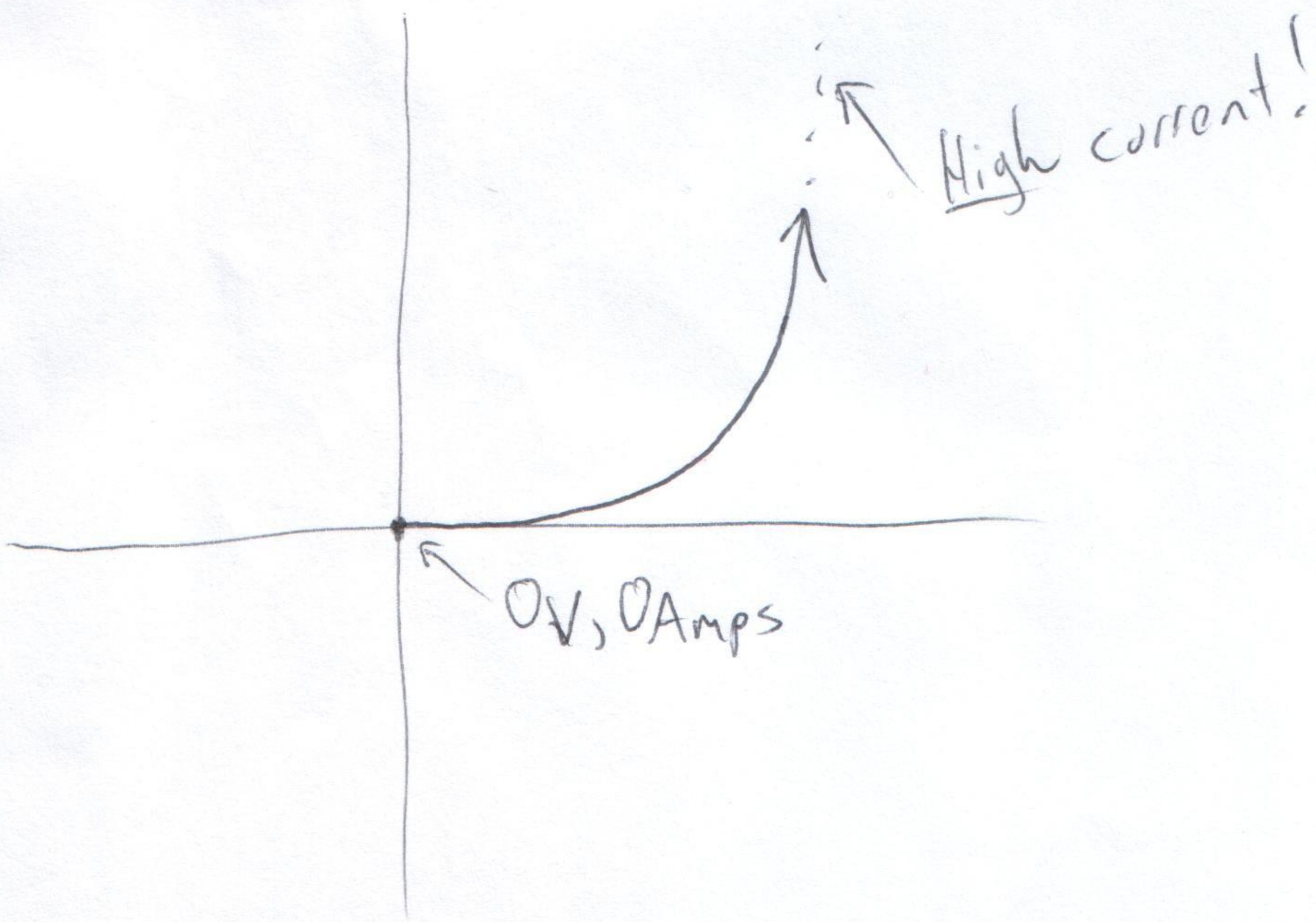


$V_{APPLIED}$
 electrons leave the p-type side for the battery, de-ionizing Acceptors.

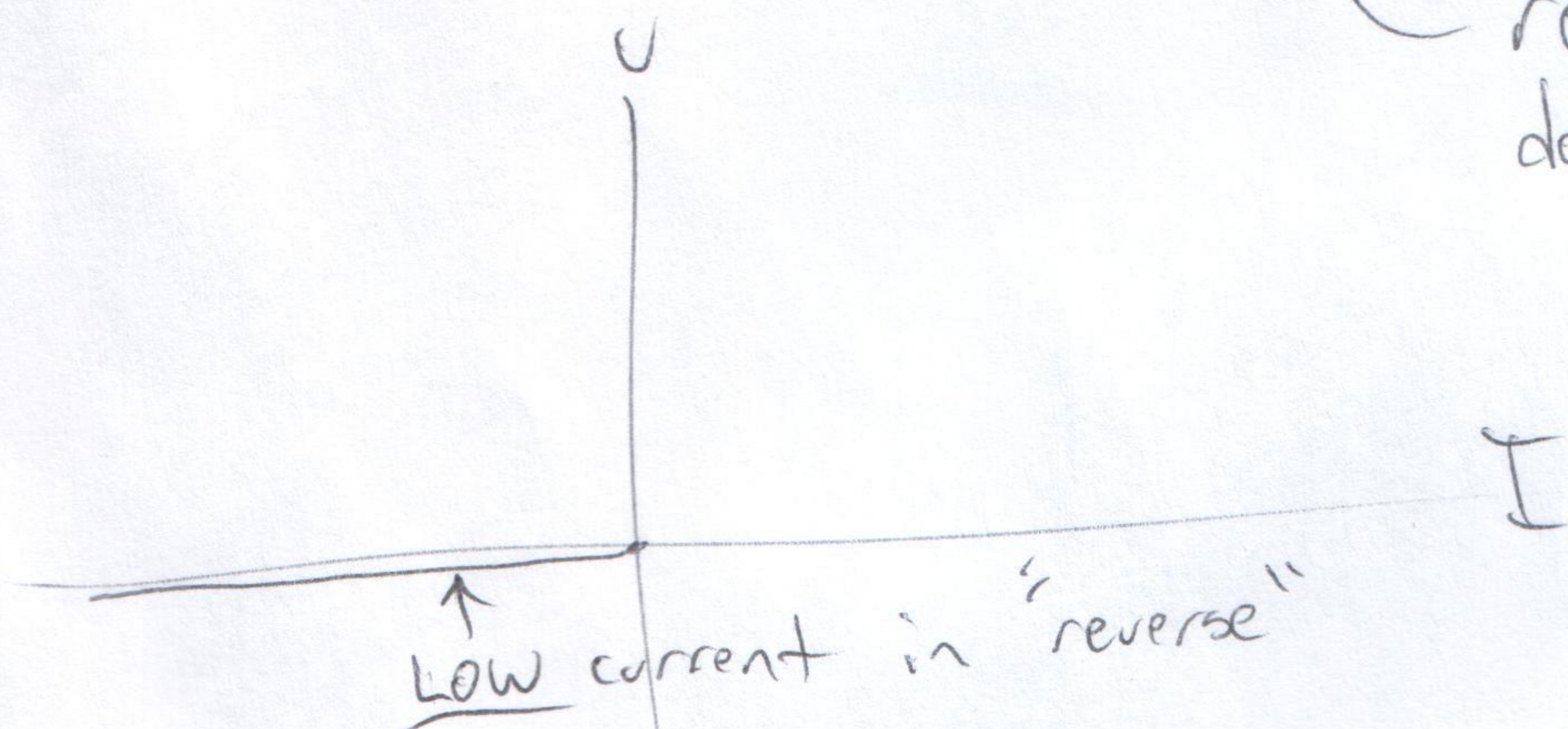
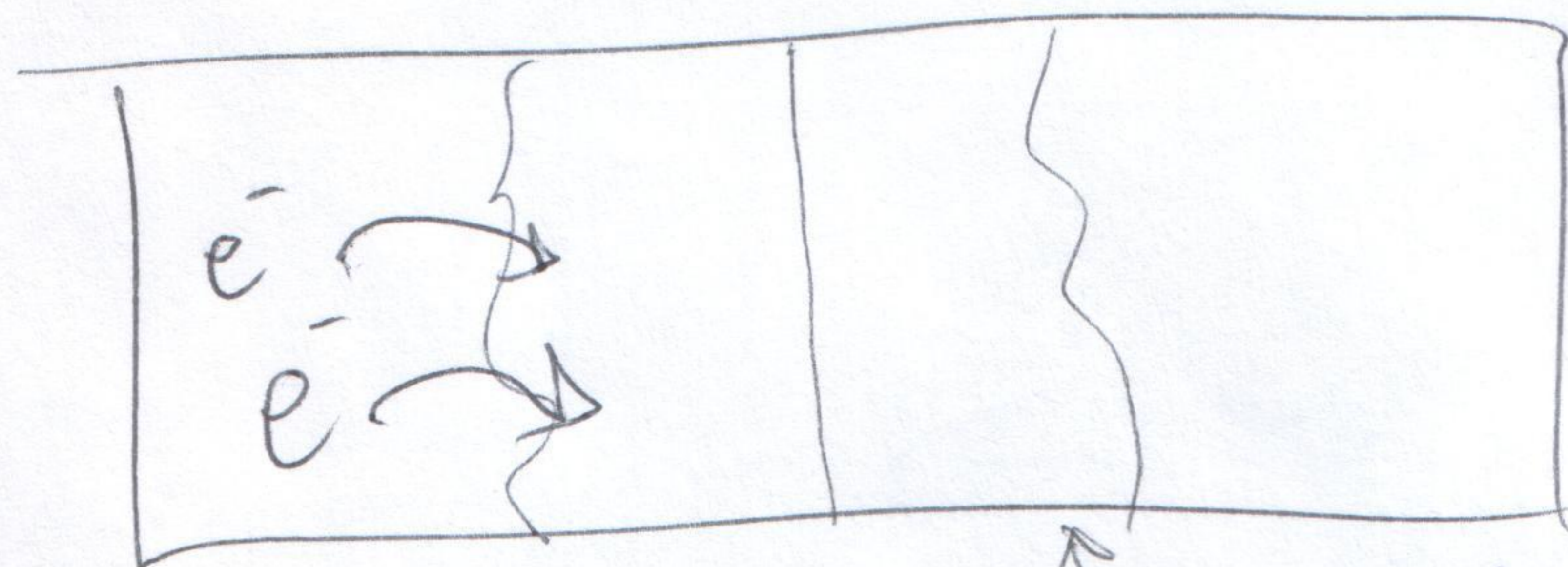
Now need fewer ionized Donors to fill the hole.

- As the "capacitor" is "discharged" (note, this doesn't mean it's zero at zero volts across it, diodes are weird) the distance covered by this region shrinks.

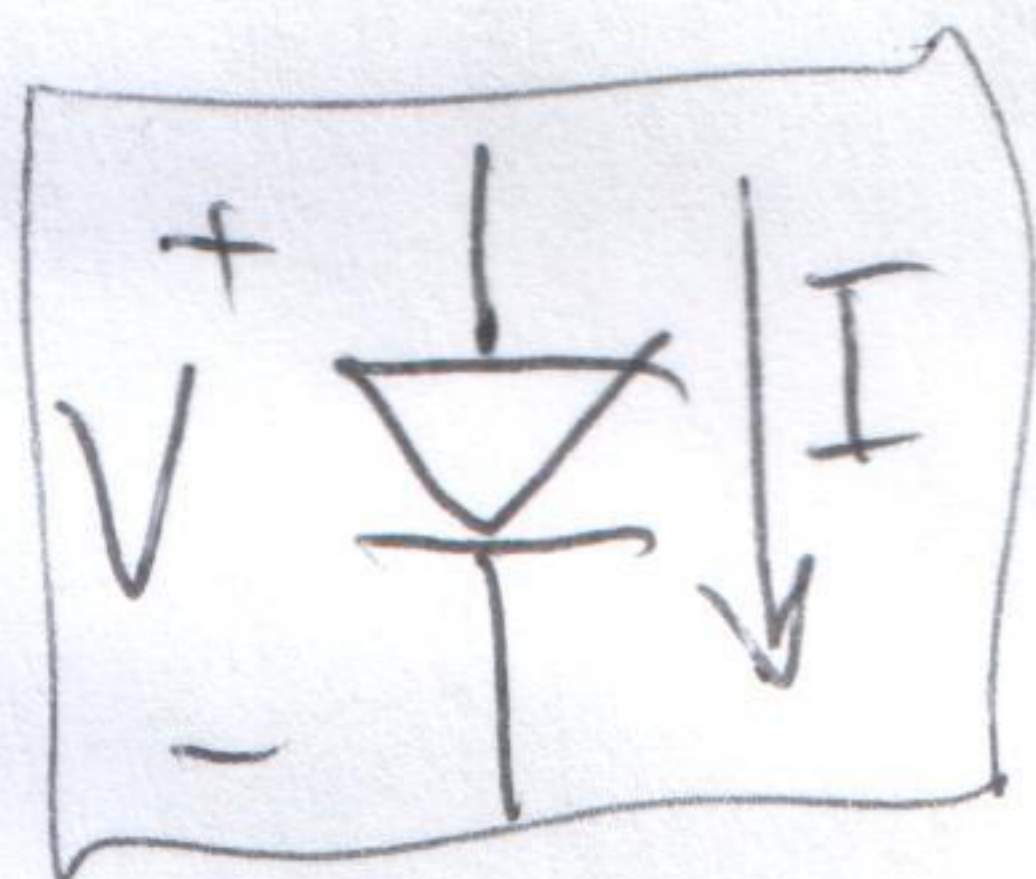
- Eventually the distance drops to zero and conduction gets even faster.



Applied Voltage in the other direction has the opposite effect as electrons stuff in on the far side:



for today, we'll leave it at this, with the following equation:

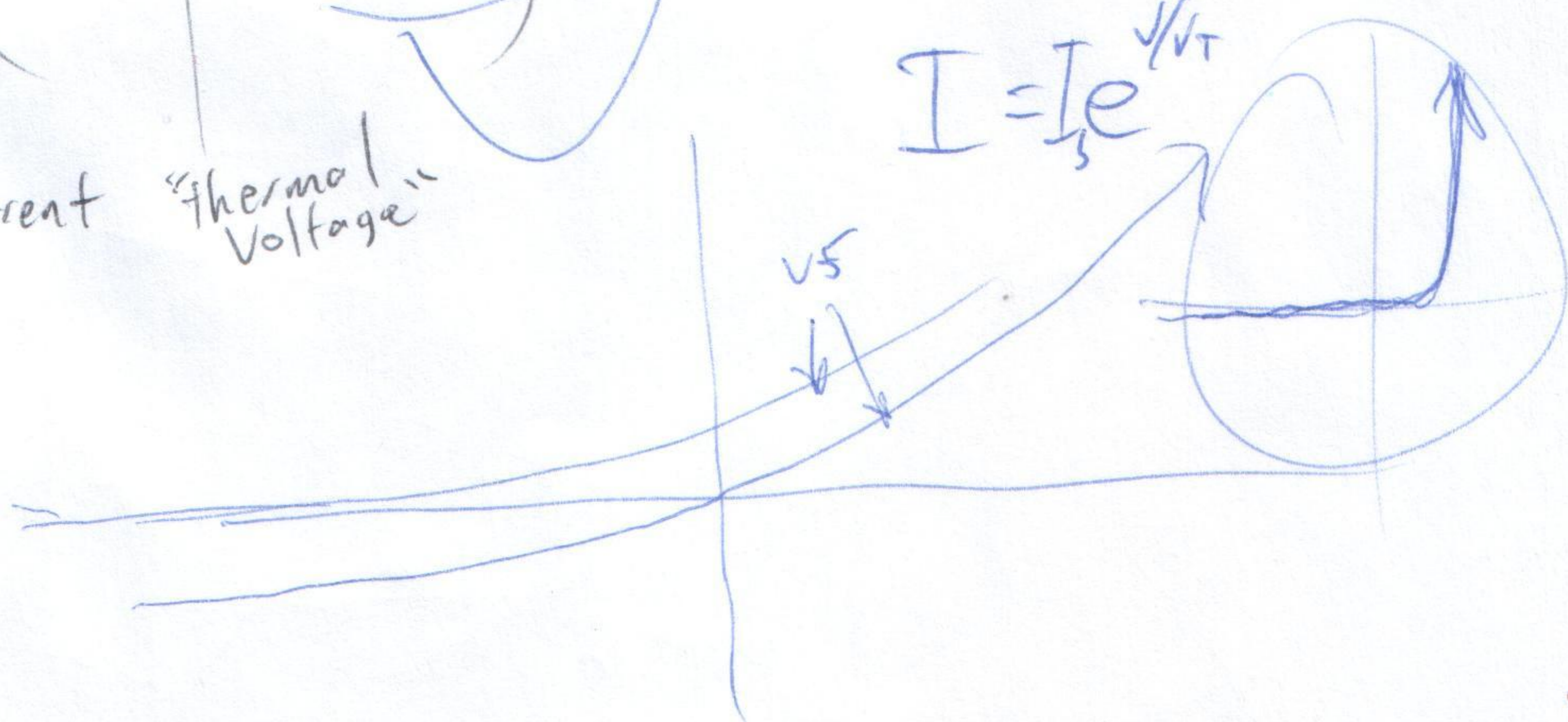


$$I = I_s \left(e^{\frac{V}{V_T}} - 1 \right)$$

const. Leakage current
Thermal Voltage

this one exists to pin $e^0 - 1 = 0$ for current, so usually we can say:

$$I = I_s e^{\frac{V}{V_T}}$$

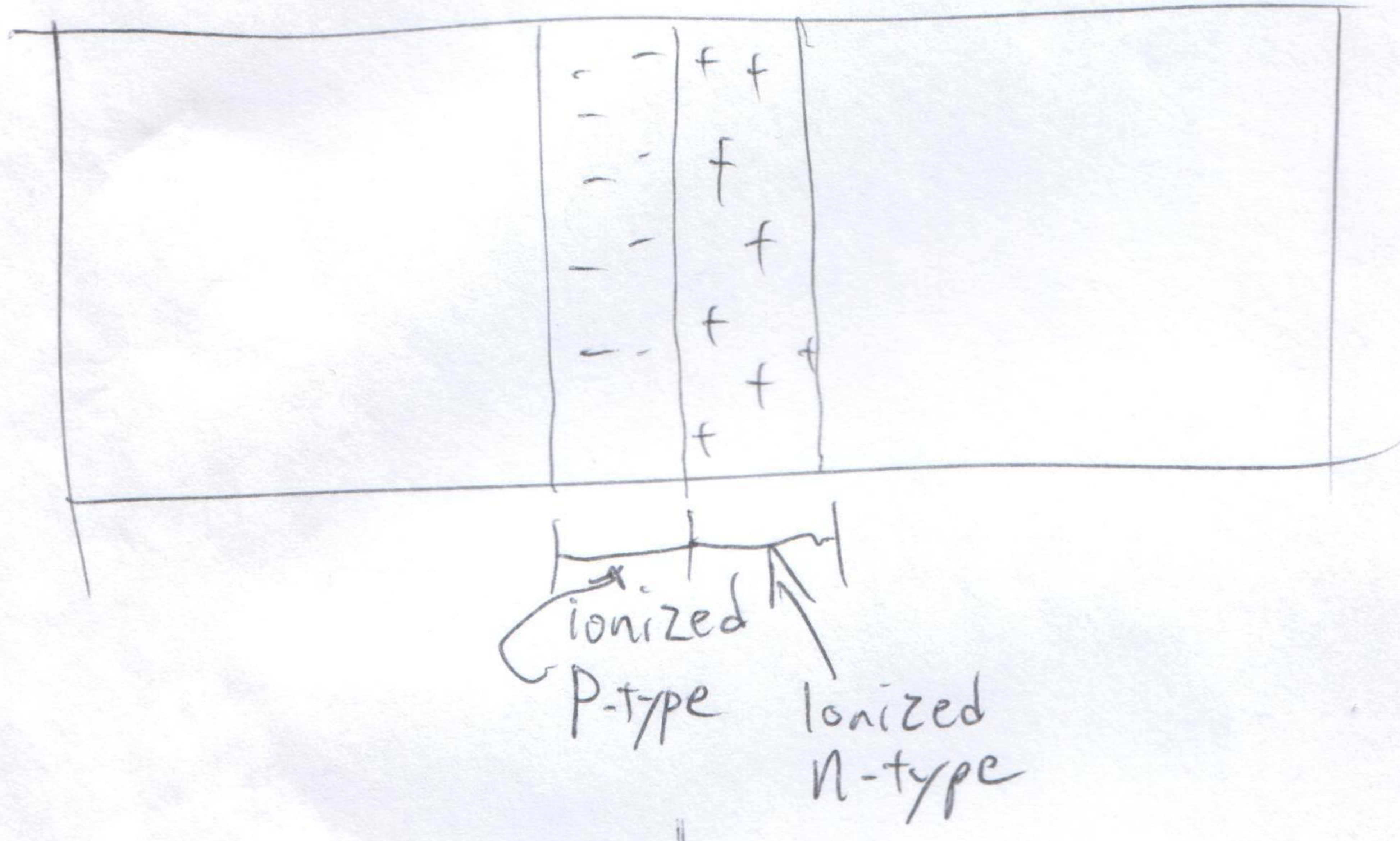


Note: V_{BI} is only not disturbed if applied voltage is also V_{BI} .

Textbook's example of a Diode:

$N_A \approx P_p = 10^{17} \text{ h/cm}^3$	$N_D \approx n_n = 10^{16} \text{ e/cm}^3$
$n_p = \frac{n_i^2}{N_A}$	$p_n = \frac{n_i^2}{N_D}$

← DIFFUSION!
→

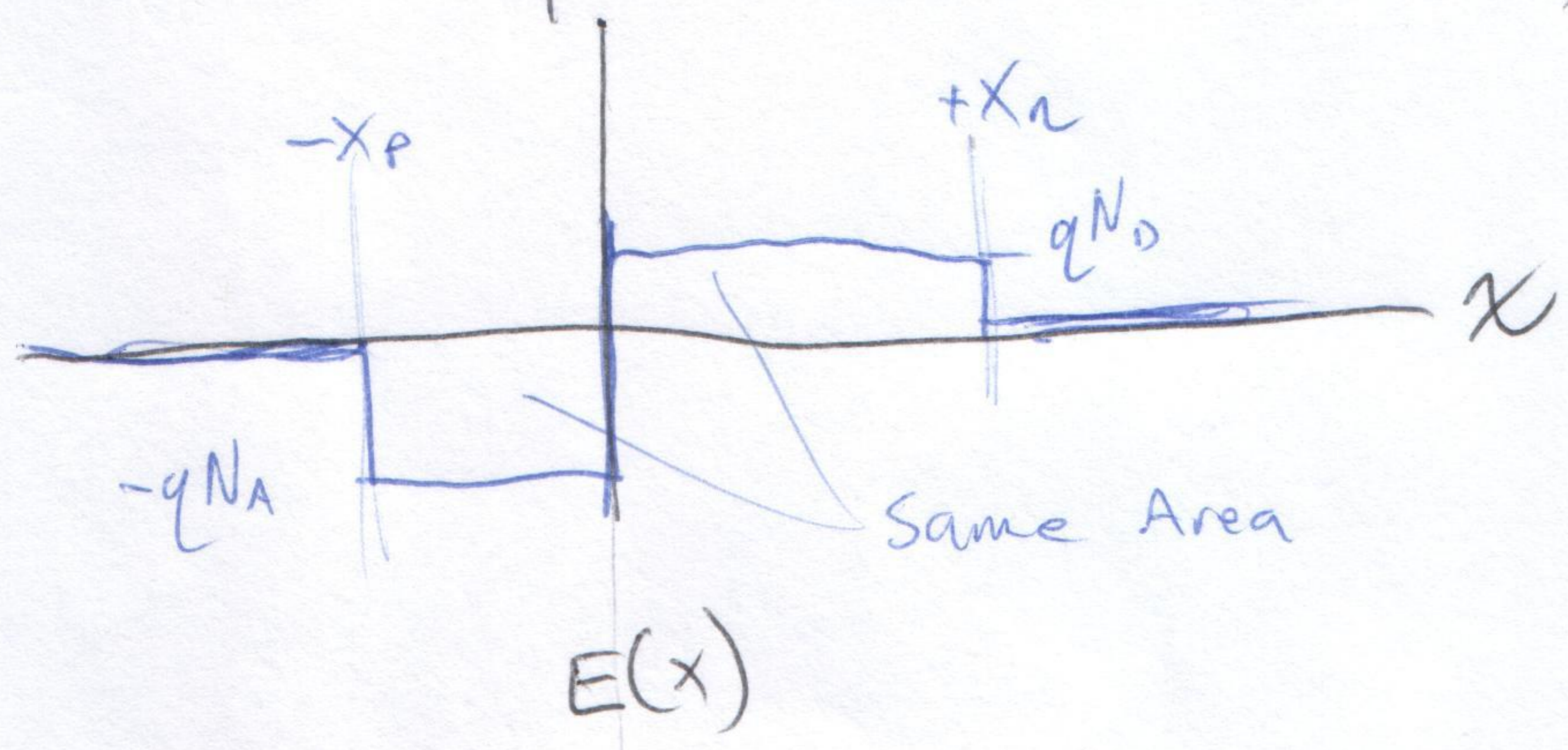


each region will have the same # of charges:

$$q N_A X_p = q N_D X_n$$

Note! This relates X_p & X_n to each other, but won't help you find X_p or X_n if you don't have at least one! **PROBLEM!**

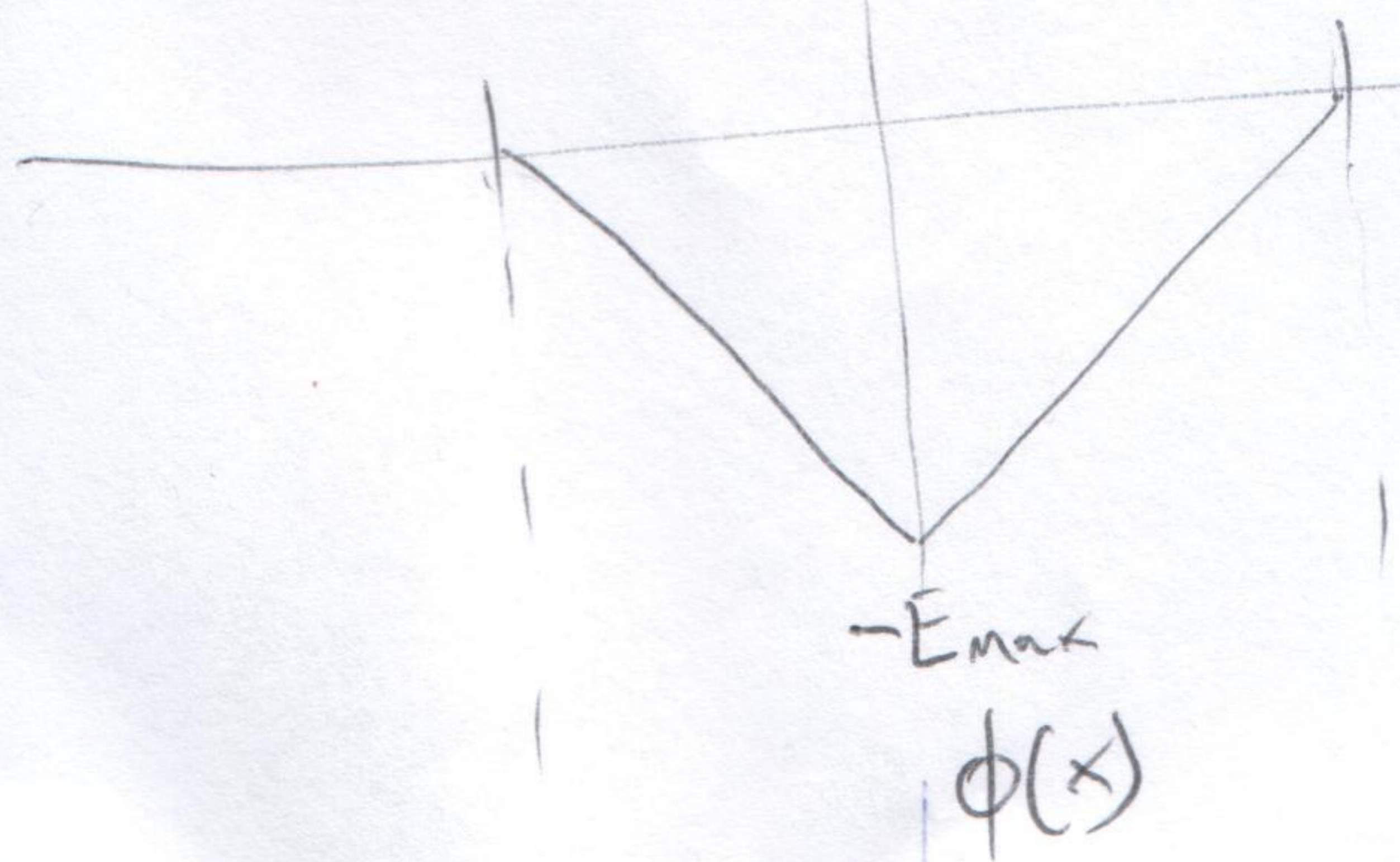
$\rho(x)$ ← charge density, C/cm^3



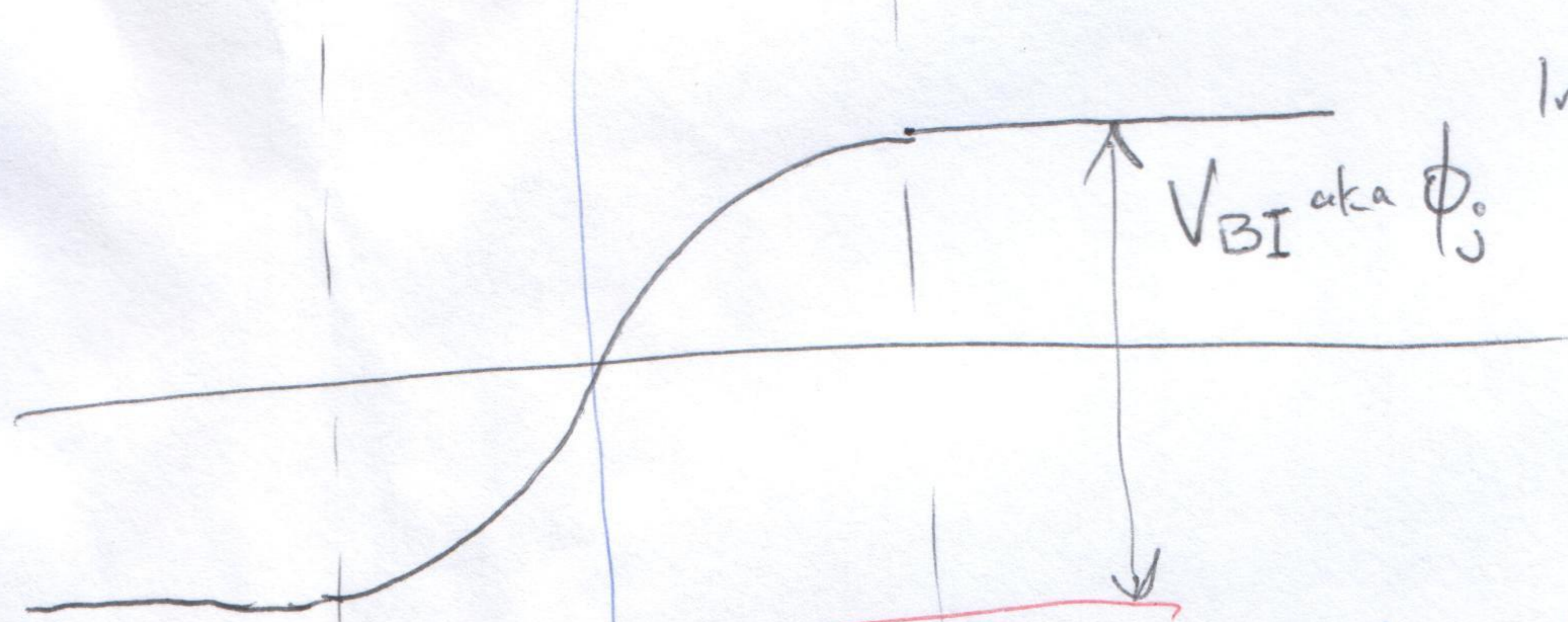
One of Maxwell's equations actually applies here: $\nabla \cdot E = \frac{\rho}{\epsilon_s}$
 In a straight line, this can reduce down to:

$$E(x) = \frac{1}{\epsilon_s} \int \rho(x) dx$$

Naturally the E-field is biggest at the interface, essentially two big ionized blobs pulling on each other



$V = \int E$] definition of voltage



Integrating twice gives V_{BI} aka ϕ_0 built-in voltage.

The ugly, awful, cheap trick that would never fly in any journal ever:

$$\phi_0 = V_T \ln \left(\frac{N_A N_D}{n_i^2} \right)$$

← pulled out of SOME OTHER BOOK with ZERO explanation!

Still troubling, but at least believable:

$$W_{do} = (x_n + x_p) = \sqrt{\frac{2\epsilon_s}{q} \left(\frac{1}{N_A} + \frac{1}{N_D} \right) \phi_0}$$

$$V_T = \frac{kT}{q}$$

Balancing Diffusion & Drift:

~~Diffusion:~~ Drift: $\vec{j} = q \cdot \# \cdot \mu \cdot \vec{E}$

Diffusion: $\vec{j} = q D \nabla \cdot \#$
gradient

$$D = \mu \frac{kT}{q}$$

Einstein's Relationship

$$j_n = 0 = q n \mu_n E + q D_n \frac{\partial n}{\partial x} = 0$$

$$nE + V_T \frac{\partial n}{\partial x} = 0 \quad] \text{ 1st order diff eq in } x$$

Labs this week: get your kits in EEB134

The website will soon be updated with lab notes!

Bring cash in case you need
- DMM - wire
- Probes
- Breadboards!

We're about to leave physics and chemistry behind for a while, but we'll be back for MOSFETs, so,

let's review key points:

— in semiconductors, conduction is DRIFT CURRENT, $\vec{j} = q \cdot \# \cdot \mu \cdot E$

— concentration gradients will cause DIFFUSION CURRENT, $\vec{j} = q \cdot \# \cdot \mu \cdot V_T \cdot \frac{\partial \#}{\partial x}$

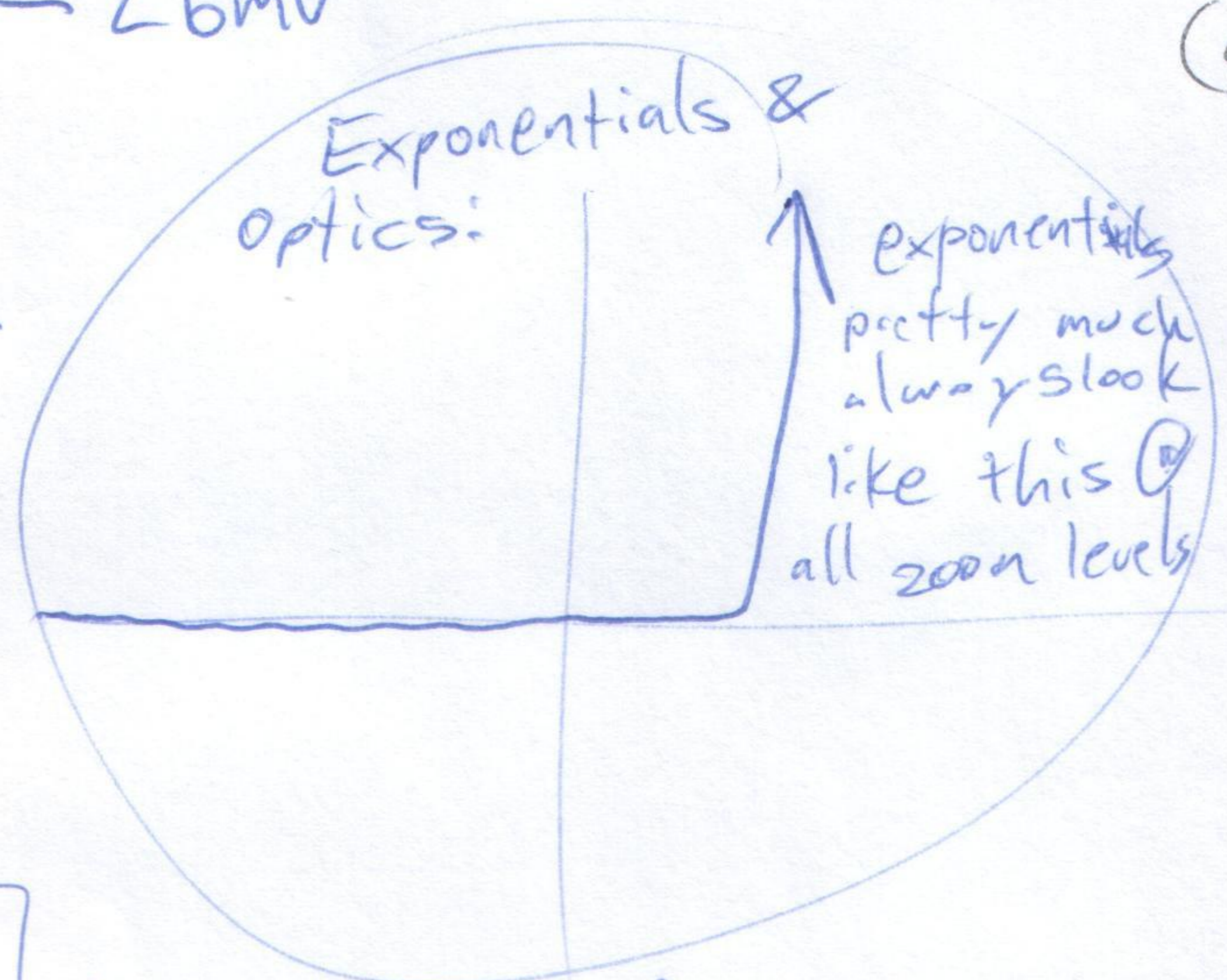
— all gradients will cause non-ohmic effects, but diode junctions especially will have exponential behavior!

ASYMMETRIC

Above all, remember that the essence of semiconductors is that CARRIER CONCENTRATIONS can be ALTERED to control CONDUCTION.

Now, circuits. $I = I_s (e^{\frac{V}{nV_T}} - 1)$

V_T in silicon: $\approx 26\text{mV}$
 $10^{-18} \leq I_s \leq 10^{-9}$
 faantastic



n : a dimensionless fudge factor (semiconductors are loaded with these.)

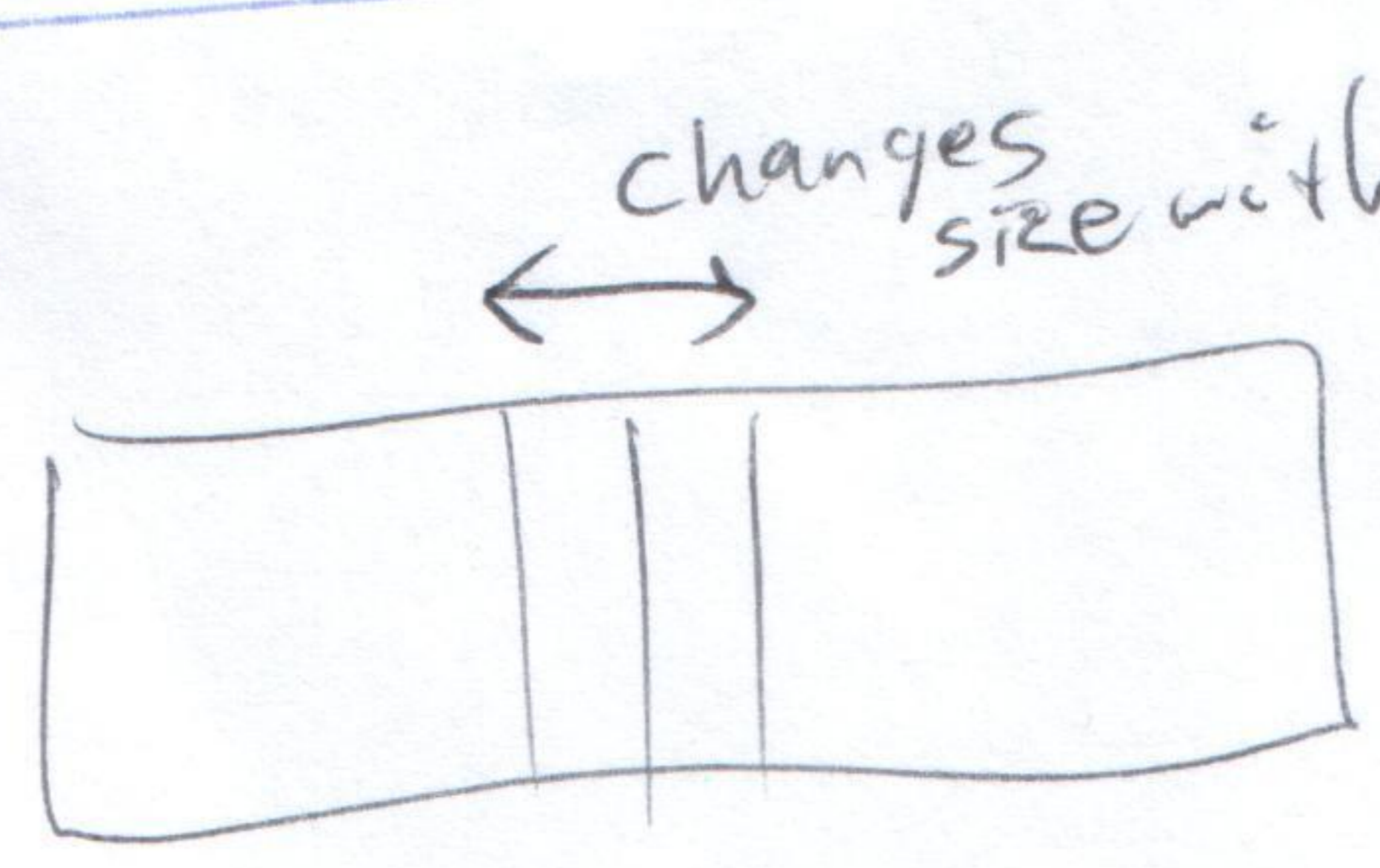
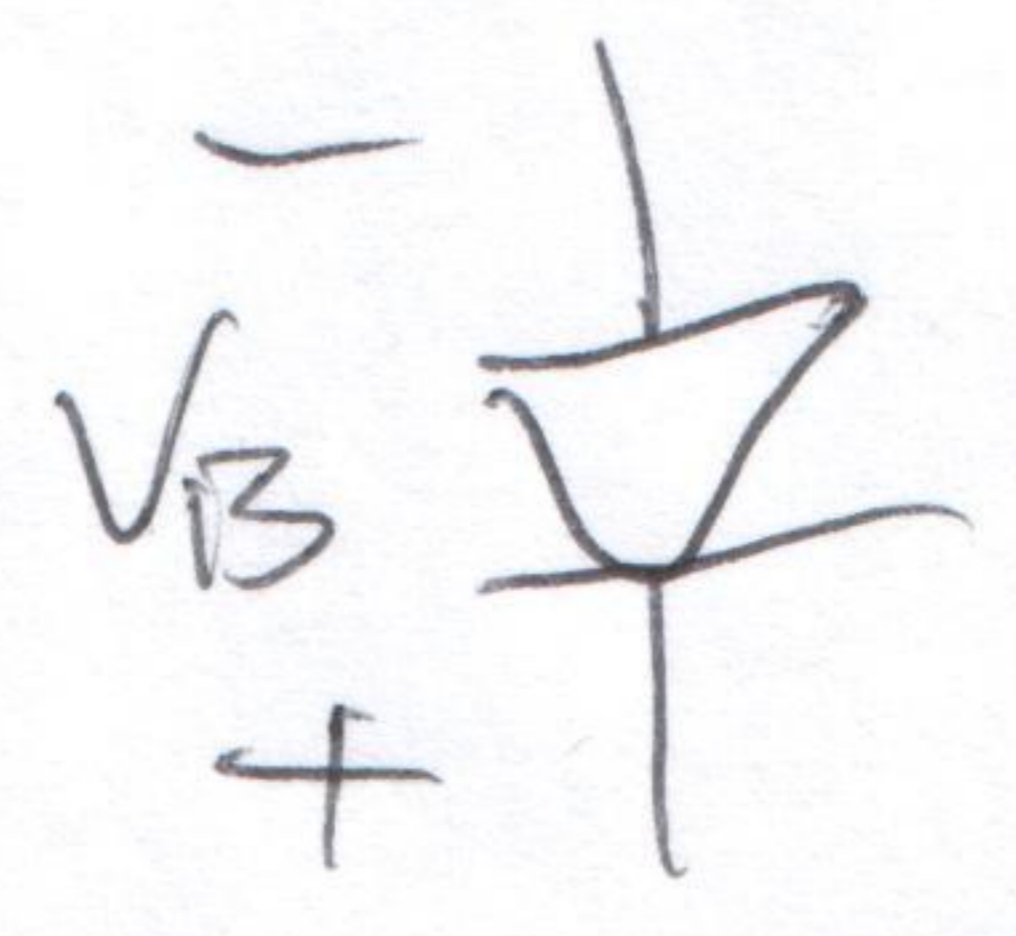
$I = I_s e^{\frac{V}{V_T}}$ (roughly true)

$\ln \frac{I}{I_s} = \frac{V}{V_T}$

$V = V_T \ln \frac{I}{I_s}$
 temperature dependent!

Also temp dependent, so unfortunately $V \neq PTAT$

In Reverse Bias



Variable $C^?$